

## Robust Control System Design for a Flexible arm by a Two-Degree-of-Freedom Compensator

Y. Shimomoto\*, T. Kobayashi\*\*, S. Miyaara\*\* and T. Ishimatsu\*

\*Mechanical Systems Engineering, Nagasaki University, Nagasaki, JAPAN

\*\* Faculty of Engineering, Kyushu Institute of Technology, Kitakyusyu, JAPAN

### ABSTRACT

This paper is concerned with a two-degree-of-freedom control system design for a flexible arm. a two-degree-of-freedom control system can achieve a robust stability specification and a control performance specification independently. By this property we improve the control performance with maintaining the same robust stability level as that of the one-degree-of-freedom control system. At First we design a two-degree-of-freedom control system which includes a feedforward controller and a feedback controller. The feedforward controller can be given by specifying a transfer function of a desired closed-loop model. We obtain a feedback controller by solving a mixed sensitivity problem. Several numerical results show that two-degree-of-freedom control systems achieve a better control performance than that of one-degree-of-freedom control systems.

### 1. INTRODUCTION

In the design for servo control systems one of the most fundamental specifications is to track the desired trajectory correctly. Even in the presence of modelling errors the servo control systems have to achieve a tracking specification. A design problem of servo control systems in the presence of modelling error is called a robust servo control system design problem.

One of the important robust control systems is a two-degree-of-freedom control system<sup>(1)</sup>. a two-degree-of-freedom control system can achieve a robust stability specification and control performance specification independently. This property indicates that a two-degree-of-freedom control system can improve the control performance with maintaining the same robust stability level as that of a one-degree-of-freedom control system.

In this report the usefulness of a two-degree-of-freedom control system is numerically demonstrated

on a flexible arm. In the control system design for a flexible arm many papers have paid attention to only robust stabilization<sup>(2), (3)</sup>. We consider not only robust stabilization but also improvement in control performances.

This report is organized as follows. In section 2 a mathematical model of the flexible arm is derived. Section 3 is devoted to the control problem setup. We formulate robust stabilization problem as a mixed sensitivity problem in which the argumented plant is constructed with frequency weighting functions. In section 4 several numerical simulations are carried out in order to evaluate the robustness and control performance of two-degree-of-freedom control system. These numerical results show that the two-degree-of-freedom control system improve the control performance with maintaining the same robust stability level as that of one-degree-of-freedom control system.

We use following notations in this report.

$\|\cdot\|_{\infty}$  denotes the  $H_{\infty}$ -norm. Let  $G(s)$  be any proper transfer function. A State-space realization of  $G(s)$  is denoted by

$$G(s)=[A, B, C, D] \quad (1)$$

### 2. PLANT DESCRIPTION

In this section a mathematical model of a flexible arm is obtained. A schematic manipulator is shown in Fig. 1. The notations in Fig. 1. are shown in Table 1. Parameters of the system are given as Table 2. The flexible arm is attached at a satellite body which is located at the origin of the coordinate. The lumped mass is attached at the tip of the arm.

By several assumptions about the arm, a mathematical model of the arm can be obtained.

$$J_r \ddot{\theta}(t) + \rho A \int_0^l xy(x,t)dx + M_c Ly(L,t) = u(t) \quad (2)$$

$$EI \frac{\partial^4 y(x,t)}{\partial t^4} + \rho A \{x\theta(t) + y(x,t)\} = 0 \quad (3)$$

where  $J_r = J_0 + \frac{\rho AL^3}{3} + M_e L^2$

The boundary conditions are given as follows.

$$EI \frac{\partial^3 y(L,t)}{\partial t^3} = M_e \{L\theta(t) + y(L,t)\} \quad (4)$$

$$\frac{\partial^2 y(L,t)}{\partial t^2} = 0 \quad (5)$$

By eigenfunctions corresponding to the boundary conditions (4) and (5), (2) and (3) can be rewritten as following infinite number of ordinary differential equations<sup>(6)</sup>.

$$q_i(t) + 2\zeta\omega_i q_i + \omega_i^2 q_i = b_i u(t) \quad (6)$$

(i = 1, 2, \dots)

where  $q_i$  is the mode coordinate of the i-th oscillation mode.  $\zeta$  denotes the damping coefficient and  $\omega_i$  denotes the natural frequency of the i-th oscillation mode.

The Angle of the arm is obtained by (7).

$$\theta(t) = b^T q \quad (7)$$

where  $q = [q_1, q_2, \dots]^T$  and  $b = [b_1, b_2, \dots]^T$ .

By (6) and (7) the transfer function from the applied torque to the angle of the arm is given as follows.

$$G(s) = \frac{A_0}{s^2} + \sum_{i=1}^{\infty} \frac{A_i}{s^2 + 2\zeta\omega_i s + \omega_i^2} \quad (8)$$

The Table 3 shows the first six parameter of the transfer function (8).

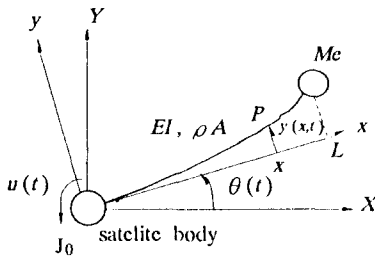


Fig.1 Schematic Diagram of the arm

Table 1 Notations in Fig. 1

$J_0$	Moment of inertia of the satellite body
$L$	Length of the arm
$\rho A$	Linear density of the arm
$EI$	Flexural rigidity
$M_e$	Lumped mass at the tip of the arm
$J_e$	Moment of inertia of the lumped mass at the tip of the arm
$u(t)$	Applied torque
$\theta(t)$	Angle of the arm

Table2 Parameter of Plant

$L$ [m]	$\rho A$ [kg/m]	$EI$ [Nm]	$M_e$ [kg]
1.0	0.77	3.4	0.1

Table 3 Parameters of transfer function

$i$	$2\zeta\omega_i$	$\omega_i^2$	$A_i$
0	0	0	2.4336
1	0.054	182.25	8.6436
2	0.1772	1962.5	1.21
3	0.492	15129	0.1444
4	0.988	61009	0.0361
5	1.66	172230	0.0121

### 3. CONTROLLER DESIGN

For the flexible arm discussed in the previous section, our control objective is to achieve a good control performance in the presence of model uncertainty. In this section control problem is formulated.

In general a two-degree-of-freedom control system is obtained as shown in Fig. 2<sup>(9)</sup>.

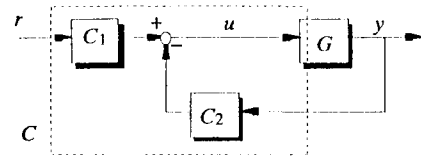


Fig. 2 Two-degree-of-freedom control system

The plant input  $u$  can be described by using the plant output  $y$  and the reference input  $r$  as follows.

$$u = C_1 r - C_2 y \quad (9)$$

where  $C_1$  represents a feedforward controller and  $C_2$  represents a feedback controller.

In this report these controllers  $C_1$  and  $C_2$  are given by (10) and (11).

$$C_1 = (G_{nom}^{-1} + C_B) G_m \quad (10)$$

$$C_2 = C_B \quad (11)$$

where  $G_{nom}$  denotes a transfer function of a nominal plant.  $G_m$  is a desired transfer function of a closed loop system, which can regard as a free parameter of the feedforward controller  $C_1$ .  $C_B$  is a robust stabilizing controller. From (10) and (11) the two-degree-of-freedom control system is constructed as shown in Fig. 3. In Fig. 3  $G_{per}$  represents the perturbed plant.

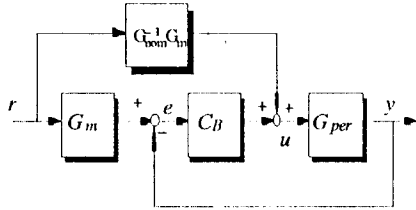


Fig. 3 Configuration of two-degree-of-freedom control system

In order to design the robust stabilizing controller  $C_B$  we solved the mixed sensitivity problem. The design specification of the mixed sensitivity problem is given by (12)<sup>(9)</sup>.

$$\left\| \begin{array}{l} W_1(s) S(s) \\ W_2(s) T(s) \end{array} \right\|_{\infty} < \gamma \quad (12)$$

where  $S(s) = (1 + G_{nom}(s)C_B(s))^{-1}$  and  $T(s) = C_B(s) (1 + G_{nom}(s)C_B(s))^{-1}$ .  $\gamma$  is a pre-specified positive number. In (12)  $W_1(s)$  and  $W_2(s)$  are weighting functions which depend on frequency.  $W_1(s)$  is usually chosen so that its gain is relatively large in a low frequency range and  $W_2(s)$  is usually chosen so that its gain is relatively large in a high frequency range. For a given nominal plant the mixed sensitivity problem is to find a controller  $C_B$  such that the closed loop system is internally stable and specification (12) is satisfied.

#### 4. NUMERICAL RESULTS

We demonstrate numerical simulations for two types of a nominal plant model. One nominal plant model includes a rigid body mode and the first two flexible modes (Case I). The other nominal plant model only includes the rigid body mode (Case II). In Fig. 4 and Fig. 5 we use the solid line to show the results of a two-degree-of-freedom control system and dashed line to show those of a one-degree-of-freedom control system.

##### (I) Case I

The weighting functions in the specification (12)

$W_1(s)$  and  $W_2(s)$  are chosen as

$$W_1(s) = \frac{0.7}{s + 0.00001} \quad (13)$$

$$W_2(s) = \frac{170s + 500}{s + 20000} \quad (14)$$

For these weighting functions the robust stabilizing controller  $C_B$  can be obtained by the  $H_{\infty}$  control theory<sup>(9)</sup>. The desired transfer function of the closed loop system  $G_m$  is given by

$$G_m(s) = \frac{s + 13}{s^3 + 4s^2 + 11.9s + 13} \quad (15)$$

From (10) we can get the feedforward controller. We apply these controllers to the flexible arm described in (8). Fig. 4 shows the step response of the closed-loop system. From Fig. 4 it can be seen that the two-degree-of-freedom control system achieves better control performance than one-degree-of-freedom control system. We also confirm that two-degree-of-freedom control system rejects effect of ignored dynamics by Fig. 4. Fig. 5 shows the plant input. As shown in Fig. 5 the plant input is oscillating. This oscillation is caused by the feedforward controller. The feedforward controller includes the flexible modes which are contained in the nominal plant model. The oscillational response of plant input is caused by the effect of these flexible modes. Next we consider removing this oscillation.

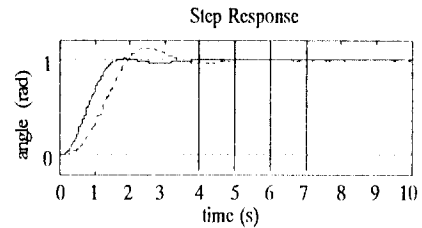


Fig. 4 Step response (Case I)

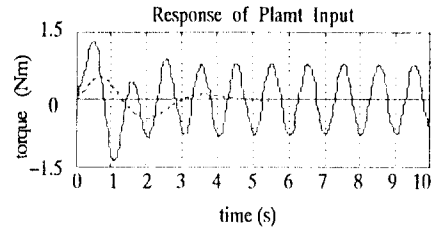


Fig. 5 The plant input (Case I)

##### (II) Case II

As mentioned before, the plant input is oscillated by the effect of flexible modes included in the

nominal plant model. In order to remove this oscillation we consider the other nominal plant model which only includes the rigid body mode. The weighting functions in the specification (12)  $W_1(s)$  and  $W_2(s)$  are chosen as

$$W_1(s) = \frac{0.85}{s + 0.00001} \quad (16)$$

$$W_2(s) = \frac{2200s^2 + 2200s + 1100}{s^2 + 200s + 20000} \quad (17)$$

The desired transfer function of the closed loop system  $G_m$  is give by

$$G_m(s) = \frac{5.5s + 10}{s^5 + 6s^4 + 20s^3 + 30s^2 + 27s + 10} \quad (18)$$

For these  $W_1(s)$ ,  $W_2(s)$  and  $G_m$  we design robust stabilizing controller  $C_B$  and feedforward controller  $C_1$ .

Fig. 6 shows the step response of the closed loop system. From Fig. 6 we can see that the two-degree-of-freedom control system rejects the effect of ignored dynamics. Fig. 7 shows the response of the plant input. As shown in Fig. 7 it can be seen that the oscillational plant input is removed.

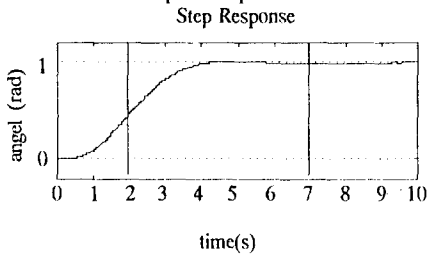


Fig. 6 Step response (Case II)

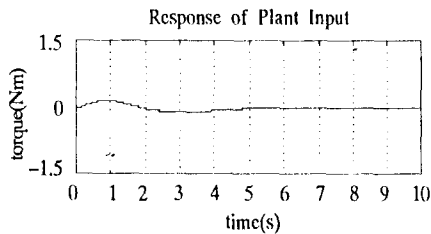


Fig. 7 Response of plant input (Case II)

## 5. CONCLUSION

In this report two-degree-of-freedom control system design for a flexible arm has been presented.

At First we consider a nominal plant model which includes a rigid body mode and the first two flexible modes. For this nominal plant model we design a two-degree-of-freedom control system. Numerical results show that the two-degree-of-

freedom control system achieves better control performance than a one-degree-of-freedom control system. However The plant input oscillates by the effect of flexible modes which are included in the nominal plant.

Secondly in order to remove the oscillation the of the plant input we consider the other nominal plant model which only includes the rigid body mode. For this nominal plant model a two-degree-of-freedom control system also achieves better control performance. Further more, the oscillational the plant input is removed.

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