

## NFL-ROO에 기준한 SMC의 안정도 증명 : Part 6

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## Stability Proof of NFL-ROO-based SMC : Part 6

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**[Abstract]** This paper presents the stability proof of a nonlinear feedback linearization-reduced order observer-based sliding mode controller (NFL-ROO-based SMC). The closed-loop stability is proved by a Lyapunov function candidate using an addition form of the sliding surface vector and the estimation error.

**Keywords :** nonlinear feedback linearization-reduced order observer-based sliding mode controller, Lyapunov function, stability proof

### 1. Introduction

In this paper, a nonlinear feedback linearization-reduced order observer-based sliding mode controller (NFL-ROO-based SMC) to design the simpler observer-based controller, and to solve the problem associated with the unmeasurable state variables in the conventional SMC is developed [1-18]. The proposed NFL-ROO-based SMC is obtained by the estimated state variable based on observer state in designing a sliding surface gain to ensure the stability by Lyapunov's method. The closed-loop stability is proved by a Lyapunov function candidate using an addition form of the sliding surface vector and the estimation error.

### 2. NFL-ROO-based SMC

The state equation for full-state feedback and the output equation based on nonlinear feedback linearization (NFL) can be expressed as [19]

$$z(t) = T(x(t)) \quad (1)$$

$$\dot{z}(t) = Az(t) + Bu(t) \quad (2)$$

$$y(t) = Cz(t) \quad (3)$$

The transformation matrix is introduced by

$$q(t) = Tz(t) \quad (4)$$

$$\begin{bmatrix} y(t) \\ q(t) \end{bmatrix} = \begin{bmatrix} C \\ T \end{bmatrix} z(t) \quad (5)$$

$$z(t) = \begin{bmatrix} C \\ T \end{bmatrix}^{-1} \begin{bmatrix} y(t) \\ q(t) \end{bmatrix} = \begin{bmatrix} P & M \end{bmatrix} \begin{bmatrix} y(t) \\ q(t) \end{bmatrix} = Py(t) + Mq(t) \quad (6)$$

The estimate  $\hat{z}$  of  $z$  is generated as

$$\hat{z}(t) = Py(t) + M\hat{q}(t) \quad (7)$$

A new realization for the system is expressed as

$$E\dot{z}(t) = EAz(t) + EBu(t) \quad (8)$$

Substituting for  $z$  and  $E$ , we get

$$\begin{aligned} E\dot{z}(t) &= \begin{bmatrix} C \\ T \end{bmatrix} \dot{z}(t) = \begin{bmatrix} \dot{y}(t) \\ \dot{q}(t) \end{bmatrix} = \begin{bmatrix} C \\ T \end{bmatrix} Az(t) + \begin{bmatrix} C \\ T \end{bmatrix} Bu(t) \\ &= \begin{bmatrix} C \\ T \end{bmatrix} A \begin{bmatrix} P & M \end{bmatrix} \begin{bmatrix} y(t) \\ q(t) \end{bmatrix} + \begin{bmatrix} C \\ T \end{bmatrix} Bu(t) \end{aligned} \quad (9)$$

Therefore, we get

$$\begin{bmatrix} \dot{y}(t) \\ \dot{q}(t) \end{bmatrix} = \begin{bmatrix} CAP & CAM \\ TAP & TAM \end{bmatrix} \begin{bmatrix} y(t) \\ q(t) \end{bmatrix} + \begin{bmatrix} CB \\ TB \end{bmatrix} u(t) \quad (10)$$

$$:= \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} y(t) \\ q(t) \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u(t) \quad (11)$$

$$\dot{y}(t) = A_{11}y(t) + A_{12}q(t) + B_1u(t) \quad (12)$$

$$\dot{q}(t) = A_{21}q(t) + A_{22}y(t) + B_2u(t) \quad (13)$$

The estimator equation for  $q$  is

$$\hat{q}(t) = A_{21}\hat{q}(t) + A_{22}y(t) + B_2u(t) + L(y(t) - C\hat{z}(t)) \quad (14)$$

Let  $CP = I$  and  $CM = 0$

$$\begin{aligned} y(t) - C\hat{z} &= y(t) - C(Py(t) + M\hat{q}) \\ &= y(t) - CPy(t) - CM\hat{q} = y(t) - y(t) - 0 = 0 \end{aligned} \quad (15)$$

The reduced-order observer is represented by

$$\begin{aligned} \dot{q}(t) &= A_{21}\hat{q}(t) + A_{22}y(t) + B_2u(t) \\ &\quad + L(\dot{y}(t) - A_{11}y(t) - B_1u(t) - A_{12}\hat{q}(t)) \end{aligned} \quad (16)$$

Define  $w$  and  $\dot{w}$ , we get

$$w(t) := \hat{q}(t) - Ly(t) \quad (17)$$

$$\dot{w}(t) := \dot{\hat{q}}(t) - L\dot{y}(t) \quad (18)$$

The final form of the reduced-order observer is

$$\begin{aligned} \dot{w}(t) &= (A_{22} - LA_{12})w(t) + [(A_{22} - LA_{12})L \\ &\quad + A_{21} - LA_{11}]y(t) + (B_2 - LB_1)u(t) \end{aligned} \quad (19)$$

$$\dot{z}(t) = Mw(t) + (P + ML)y(t) \quad (20)$$

$$\dot{\hat{z}}(t) = M\dot{w}(t) + (P + ML)\dot{y}(t) \quad (21)$$

The equations (19) and (20) are simplified as [1]

$$\dot{w}(t) = Fw(t) + Gy(t) + Hu(t) \quad (22)$$

$$\dot{z}(t) = Mw(t) + Ny(t) \quad (23)$$

$$\text{where, } F := (A_{22} - LA_{12}) \quad (24)$$

$$G := [(A_{22} - LA_{12})L + A_{21} - LA_{11}] \quad (25)$$

$$H := (B_2 - LB_1) \quad (26)$$

$$N := (P + ML) \quad (27)$$

The sliding surface vector and the differential sliding surface vector based on NFL are

$$\hat{z} = T^{-1}\hat{q} \quad (28)$$

$$\sigma(\hat{z}(t)) = G_{ss}^T \hat{z}(t) \quad (28)$$

$$\dot{\sigma}(\hat{z}(t)) = G_{ss}^T \dot{\hat{z}}(t) \quad (29)$$

The Lyapunov's function is

$$V(\hat{z}(t)) = \sigma^2(\hat{z}(t))/2 \quad (30)$$

The time derivative of  $V(\hat{z}(t))$  is expressed as

$$\begin{aligned} \dot{V}(\hat{z}(t)) &= \sigma(\hat{z}(t))\dot{\sigma}(\hat{z}(t)) \\ &= G_{ss}^T \dot{\hat{z}}(t) G_{ss}^T \dot{\hat{z}}(t) = G_{ss}^T \hat{z}(t) G_{ss}^T [Mw(t) + (P + ML)y(t)] \\ &= G_{ss}^T \hat{z}(t) G_{ss}^T [M((A_{22} - LA_{12})w(t) \\ &\quad + (A_{22} - LA_{12})L + A_{21} - LA_{11})y(t) \\ &\quad + (B_2 - LB_1)u_{ROO-SMC}(t)) + (P + ML) \\ &\quad ((A_{11}y(t) + A_{12}q(t) + B_1u_{ROO-SMC}(t))] \\ &= G_{ss}^T \hat{z}(t) [G_{ss}^T M(A_{22} - LA_{12})w(t) \\ &\quad + (G_{ss}^T M(A_{22} - LA_{12})L + G_{ss}^T MA_{21} \\ &\quad - G_{ss}^T MLA_{11} + G_{ss}^T (P + ML)A_{11})y(t) \\ &\quad + (P + ML)A_{12}q(t) + (MB_2 + PB_1)u_{ROO-SMC}(t)] \leq 0 \quad (32) \end{aligned}$$

The control inputs with switching function are

$$\begin{aligned} u_{ROO-SMC}(t) &\geq -[G_{ss}^T (MB_1 + PB_1)]^{-1} [G_{ss}^T (MA_{22} - MLA_{12})w(t) \\ &\quad + G_{ss}^T (M(A_{22} - LA_{12})L + MA_{21} + PA_{11})y(t) \\ &\quad + G_{ss}^T (P + ML)A_{12}q(t)] \quad \text{for } G_{ss}^T \hat{z}(t) > 0 \quad (33) \end{aligned}$$

$$\begin{aligned} u_{ROO-SMC}(t) &\leq -[G_{ss}^T (MB_1 + PB_1)]^{-1} [G_{ss}^T (MA_{22} - MLA_{12})w(t) \\ &\quad + G_{ss}^T (M(A_{22} - LA_{12})L + MA_{21} + PA_{11})y(t) \\ &\quad + G_{ss}^T (P + ML)A_{12}q(t)] \quad \text{for } G_{ss}^T \hat{z}(t) < 0 \quad (34) \end{aligned}$$

The estimated control input with sign function is

$$\hat{u}_{ROO-SMC}^{sgn}(t) = -[G_{ss}^T (MB_1 + PB_1)]^{-1} [G_{ss}^T (MA_{22} - MLA_{12})w(t)$$

$$\begin{aligned} &\quad + G_{ss}^T (M(A_{22} - LA_{12})L + MA_{21} + PA_{11})y(t) \\ &\quad + G_{ss}^T (P + ML)A_{12}q(t)] \operatorname{sign}(\sigma(\hat{z}(t))) \end{aligned} \quad (35)$$

The equation (35) is simplified as follows

$$\begin{aligned} \hat{u}_{ROO-SMC}^{sgn}(t) &= -[RGK_1 w(t) + RGK_2 y(t) + RGK_3 q(t)] \\ &\quad \operatorname{sign}(\sigma(\hat{z}(t))) \end{aligned} \quad (36)$$

where,

$$RGK_1 := [G_{ss}^T (MB_1 + PB_1)]^{-1} G_{ss}^T M(A_{22} - LA_{12}) \quad (37)$$

$$\begin{aligned} RGK_2 &:= [G_{ss}^T (MB_1 + PB_1)]^{-1} \\ &\quad G_{ss}^T (M(A_{22} - LA_{12})L + MA_{21} + PA_{11}) \end{aligned} \quad (38)$$

$$RGK_3 := [G_{ss}^T (MB_1 + PB_1)]^{-1} G_{ss}^T (P + ML)A_{12} \quad (39)$$

**Theorem 1:** Consider the full state equation based on NFL for the regulation problem and the reduced order observer problem based on NFL

$$\dot{z} = Az + Bu_{ROO-SMC}^{sgn} \text{ and } y = Cz$$

$$\dot{z} = M\dot{w} + (P + ML)\dot{y}$$

$$\begin{aligned} \dot{w} &= (A_{22} - LA_{12})w + [(A_{22} - LA_{12})L + A_{21} - LA_{11}]y \\ &\quad + (B_2 - LB_1)\hat{u}_{ROO-SMC}^{sgn} \\ \dot{y} &= A_{11}y + A_{12}q + B_1\hat{u}_{ROO-SMC}^{sgn} \\ \hat{z} &= Mw + (P + ML)y \end{aligned}$$

Suppose that  $(A, C)$  is detectable and  $(A - LC)$  is Hurwitz. The estimated sliding mode reduced order control law with sign function is guaranteed an *asymptotically stable* for the system (2)

$$\hat{u}_{ROO-SMC}^{sgn}(t) = -[RGK_1 w + RGK_2 y + RGK_3 q] \operatorname{sign}(\sigma(\hat{z}))$$

where,

$$RGK_1 := [G_{ss}^T (MB_1 + PB_1)]^{-1} G_{ss}^T (MA_{22} - MLA_{12})$$

$$\begin{aligned} RGK_2 &:= [G_{ss}^T (MB_1 + PB_1)]^{-1} \\ &\quad G_{ss}^T [M(A_{22} - LA_{12})L + MA_{21} + PA_{11}] \end{aligned}$$

$$RGK_3 := [G_{ss}^T (MB_1 + PB_1)]^{-1} G_{ss}^T (P + ML)A_{12}$$

$$\begin{aligned} \text{subject to } \operatorname{sign}(\sigma(\hat{z})) &= 1 & \text{for } \sigma(\hat{z}) > 0 \\ \operatorname{sign}(\sigma(\hat{z})) &= 0 & \text{for } \sigma(\hat{z}) = 0 \\ \operatorname{sign}(\sigma(\hat{z})) &= -1 & \text{for } \sigma(\hat{z}) < 0 \end{aligned}$$

**Proof.** Let us define the estimation error and the differential estimation error

$$e = z - \hat{z}$$

$$\begin{aligned} \dot{e} &= \dot{q} - \dot{\hat{q}} = A_{22}q + A_{21}y + B_2\hat{u}_{ROO-SMC}^{sgn} - (A_{22}\hat{q} + A_{21}y + B_2\hat{u}_{ROO-SMC}^{sgn} \\ &\quad + L(\dot{y} - A_{11}y - B_1\hat{u}_{ROO-SMC}^{sgn} - A_{12}\hat{q})) \\ &= (A_{22} - LA_{12})(q - \hat{q}) = (A_{22} - LA_{12})e \end{aligned}$$

Lyapunov's function candidate is chosen by

$$V = \frac{1}{2} \sigma^T \sigma + \frac{1}{2} e^T e$$

The derivative of a Lyapunov's function is

$$\begin{aligned}\dot{V} &= \sigma^T \dot{\sigma} + e^T \dot{e} = \sigma^T \left( G_{ss}^T \dot{z} \right) + e^T \left( A_{22} - LA_{12} \right) e \\ &= \sigma^T \left( G_{ss}^T \left( M\dot{w} + (P + ML)\dot{y} \right) \right) + e^T \left( A_{22} - LA_{12} \right) e \\ &= \sigma^T \left( G_{ss}^T M \left( A_{22} - LA_{12} \right) w + G_{ss}^T \left( (MA_{22} - MLA_{12})L \right. \right. \\ &\quad \left. \left. + (MA_{21} + PA_{11}) \right) y + G_{ss}^T (P + ML) A_{12} q \right. \\ &\quad \left. - \left( G_{ss}^T MB_2 + G_{ss}^T PB_1 \right) \left( RGK_1 w + RGK_2 y \right. \right. \\ &\quad \left. \left. + RGK_3 q \right) sign(\sigma(\hat{z})) \right) + e^T \left( A_{22} - LA_{12} \right) e \\ V' &= \sigma^T \left[ G_{ss}^T M \left( A_{22} - LA_{12} \right) w + G_{ss}^T \left( (MA_{22} - MLA_{12})L \right. \right. \\ &\quad \left. \left. + (MA_{21} + PA_{11}) \right) y + G_{ss}^T (P + ML) A_{12} q \right. \\ &\quad \left. - \left( G_{ss}^T MB_2 + G_{ss}^T PB_1 \right) \left[ G_{ss}^T \left( MB_2 + PB_1 \right) \right]^{-1} \right. \\ &\quad \left. G_{ss}^T \left( MA_{22} - MLA_{12} \right) w sign(\sigma(\hat{z})) \right. \\ &\quad \left. - \left( G_{ss}^T MB_2 + G_{ss}^T PB_1 \right) \left[ G_{ss}^T \left( MB_2 + PB_1 \right) \right]^{-1} \right. \\ &\quad \left. G_{ss}^T \left[ M \left( A_{22} - LA_{12} \right) L + MA_{21} + PA_{11} \right] y(t) sign(\sigma(\hat{z})) \right. \\ &\quad \left. - \left( G_{ss}^T MB_2 + G_{ss}^T PB_1 \right) \left[ G_{ss}^T \left( MB_2 + PB_1 \right) \right]^{-1} \right. \\ &\quad \left. G_{ss}^T \left( P + ML \right) A_{12} q sign(\sigma(\hat{z})) \right] + e^T \left( A_{22} - LA_{12} \right) e\end{aligned}$$

$$\text{Let } G_{ss}^T \left( MB_2 + PB_1 \right) \left[ G_{ss}^T \left( MB_2 + PB_1 \right) \right]^{-1} = I,$$

$$\begin{aligned}\dot{V} &= \sigma^T \left[ G_{ss}^T M \left( A_{22} - LA_{12} \right) w \right. \\ &\quad \left. + G_{ss}^T \left( (MA_{22} - MLA_{12})L + (MA_{21} + PA_{11}) \right) y \right. \\ &\quad \left. + G_{ss}^T (P + ML) A_{12} q - G_{ss}^T \left( MA_{22} - MLA_{12} \right) w sign(\sigma(\hat{z})) \right. \\ &\quad \left. - G_{ss}^T \left[ M \left( A_{22} - LA_{12} \right) L + MA_{21} + PA_{11} \right] y sign(\sigma(\hat{z})) \right. \\ &\quad \left. - G_{ss}^T \left( P + ML \right) A_{12} q sign(\sigma(\hat{z})) \right] + e^T \left( A_{22} - LA_{12} \right) e \\ &= \sigma^T G_{ss}^T M \left( A_{22} - LA_{12} \right) w \left( 1 - sign(\sigma(\hat{z})) \right) \\ &\quad + \sigma^T G_{ss}^T \left( P + ML \right) A_{12} q \left( 1 - sign(\sigma(\hat{z})) \right) \\ &\quad + \sigma^T G_{ss}^T \left( M \left( A_{22} - LA_{12} \right) L + MA_{21} + PA_{11} \right) y \left( 1 - sign(\sigma(\hat{z})) \right) \\ &\quad + e^T \left( A_{22} - LA_{12} \right) e\end{aligned}$$

If  $(A_{22} - LA_{12})$  is stable, the estimation error is  $e \rightarrow 0$  as  $t \rightarrow 0$ .

$$\begin{aligned}\dot{V} &= \sigma^T G_{ss}^T M \left( A_{22} - LA_{12} \right) w \left( 1 - sign(\sigma(\hat{z})) \right) + \sigma^T G_{ss}^T \left( P + ML \right) \\ &\quad A_{12} q \left( 1 - sign(\sigma(\hat{z})) \right) + \sigma^T G_{ss}^T \left( M \left( A_{22} - LA_{12} \right) L \right. \\ &\quad \left. + MA_{21} + PA_{11} \right) y \left( 1 - sign(\sigma(\hat{z})) \right) \leq 0\end{aligned}$$

subject to, if  $\sigma > 0$ ,  $\dot{V} = 0$

if  $\sigma = 0$ ,  $\dot{V} = 0$

if  $\sigma < 0$ ,  $\dot{V} = 2kG_{ss}^T M \left( A_{22} - LA_{12} \right) w$

$+ 2kG_{ss}^T \left( P + ML \right) A_{12} q$

$+ 2kG_{ss}^T \left( M \left( A_{22} - LA_{12} \right) L + MA_{21} + PA_{11} \right) y < 0$

where,  $k$  is positive constant. The above condition is satisfied on negative definite, and is *asymptotically stable*. This completes the proof of this theorem.  $\square$

### 3. Conclusion

A nonlinear feedback linearization-reduced order observer-based sliding mode controller (NFL-ROO-based SMC) has been proposed and a stability proof of a closed-loop system has been done.

### References

- [1] D. G. Luenberger, "Observing the state of a linear system", IEEE Trans. Mil. Electron., Vol. MIL-8, pp. 74-80, Apr. 1964.
- [2] V. I. Utkin, "Variable structure systems with sliding modes", IEEE Trans. on Automatic Control, AC-22, No. 2, pp. 212-222, April, 1977.
- [3] W. C. Chan and Y. Y. Hsu, "An optimal variable structure stabilizer for power system stabilization", IEEE Trans. on Power Apparatus and Systems, Vol. PAS-102, pp. 1738-1746, Jun., 1983.
- [4] J. J. Lee, "Optimal multidimensional variable structure controller for multi-interconnected power system", KIEE Trans., Vol. 38, No. 9, pp. 671-683, Sep., 1989.
- [5] M. L. Kothari, J. Nanda and K. Bhattacharya, 'Design of variable structure power system stabilizers with desired eigenvalues in the sliding mode', IEE Proc. C, Vol. 140, No. 4, pp. 263-268, 1993.
- [6] S. S. Lee, J. K. Park and J. J. Lee, "Sliding mode-MFAC power system stabilizer", Jour. of KIEE, Vol. 5, No. 1, pp. 1-7, Mar., 1992.
- [7] S. S. Lee and J. K. Park, "Sliding mode-model following power system stabilizer including closed-loop feedback", Jour. of KIEE, Vol. 9, No. 3, pp. 132-138, Sep., 1996.
- [8] S. S. Lee, J. K. Park et al., "Multimachine stabilizer using sliding mode-model following including closed-loop feedback", Jour. of KIEE, Vol. 9, No. 4, pp. 191-197, Dec., 1996.
- [9] S. S. Lee and J. K. Park, "Sliding mode power system stabilizer based on LQR : Part I", Jour. of EEIS, Vol. 1, No. 3, pp. 32-38, 1996.
- [10] S. S. Lee and J. K. Park, "Sliding mode observer power system stabilizer based on linear full-order observer : Part II", Jour. of EEIS, Vol. 1, No. 3, pp. 39-45, 1996.
- [11] S. S. Lee and J. K. Park, "Full-order observer-based sliding mode power system stabilizer with desired eigenvalue-assignment for unmeasurable state variables", Jour. of EEIS, Vol. 2, No. 2, pp. 36-42, 1997.
- [12] S. S. Lee and J. K. Park, "New sliding mode observer-model following power system stabilizer including CLF for unmeasurable state variables", Jour. of EEIS, Vol. 2, No. 3, pp. 88-94, 1997.
- [13] S. S. Lee and J. K. Park, "Multimachine stabilizer using sliding mode observer-model following including CLF for unmeasurable state variables", Jour. of EEIS, Vol. 2, No. 4, pp. 53-58, 1997.
- [14] S. S. Lee and J. K. Park, " $H_\infty$  observer-based sliding mode power system stabilizer for unmeasurable state variables", Jour. of EEIS, Vol. 2, No. 1, pp. 70-76, 1997.
- [15] S. S. Lee and J. K. Park, "Nonlinear feedback linearization-full order observer/sliding mode controller design for improving transient stability in a power system", Jour. of EEIS, Vol. 3, No. 2, pp. 184-192, 1998.
- [16] S. S. Lee and J. K. Park, "Nonlinear feedback linearization- $H_\infty$ /sliding mode controller design for improving transient stability in a power system", Jour. of EEIS, Vol. 3, No. 2, pp. 193-201, 1998.
- [17] S. S. Lee and J. K. Park, "Design of power system stabilizer using observer/sliding mode, observer/sliding mode-model following and  $H_\infty$ /sliding mode controllers for small-signal stability study", Inter. Jour. of Electrical Power & Energy Systems, accepted, 1998.
- [18] S. S. Lee and J. K. Park, "Design of reduced-order observer-based variable structure power system stabilizer for unmeasurable state variables", IEE PROC.-GEN., TRANS. AND DISTRIB., accepted, 1998.
- [19] R. Marino and P. Tomei, "Nonlinear control design", Prentice-Hall Press, 1995.