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4x4 매트릭스 광스위치의 최적 설계

(An Optimal Design of 4x4 Optical Matrix Switch)

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요 약

광손실의 측면에서 단순 트리구조를 갖는 GaAs/AlGaAs 매트릭스 광스위치의 최적 설계과정이 제시되었다. 도핑에 의한 광손실을 고려하여 1.3 µm의 파장에서 광손실이 0.537 dB/cm인 pin형 GaAs/AlGaAs 광도파로 기판이 설계되었다. 설계된 기판상의 rib형 광도파로를 사용하는 경우에 대해. 4x4 매트릭스 광스위치의 연결소자인 역 4b형 전기광학 스위치의 동작전압, 소자길이 및 각 연결소자간의 연결을 위한 광도파로의 굴곡손실이 광도파로 변수 및 곡률반경의 함수로 계산되었다. 광도파로 변수들 및 광도파로의 곡률반경을 고려하여 4x4 매트릭스 광스위치의 광손실을 계산한 결과, 일정한 광도파로 변수에 대해 광손실을 최소로 하는 최적의 광도파로 곡률반경이, 존재한다. 공정에 둔감하고 적은 광손실을 갖는 4x4 매트릭스 광스위치의 제작을 위해서는 rib형 광도파로의 rib 높이 및 너비가 큰 등의 광구속이 큰 조건으로 광도파로를 제작해야 할 것으로 분석되었다. 이러한 분석에 근거하여 3 dB 광손실을 갖는 동작전압 12 volt의 4x4 매트릭스 광스위치가 설계되었다.

Abstract

The design procedure of a GaAs/AlGaAs semiconductor matrix optical switch is presented for a simplified tree architecture in the viewpoint of optical loss. A low loss, 0.537 dB/cm, pin type substrate is designed by considering the loss due to impurity doping at 1.3 μ m wavelength. The operating voltage and the device length of a reversed ΔB electro-optic directional coupler(EODC) switch which is a cross-point device of the 4x4 matrix optical switch and the bending loss of rib waveguide are caculated as functions of waveguide parameters and bending parameters. There is an optimum bending radius for some waveguide parameters. It is recommend that higher optical confinement conditions such as wide waveguide width and higher rib-height should be chosen for structural parameters of a low loss and a process insensitive 4x4 matrix optical switch. A 4x4 optical matrix switch which has a 3 dB loss and a 12 volt operating voltage is designed.

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I. Introduction

There is an increasing demand for optical fiber communications, which make broadband communication networks possible. With its wide bandwidth, especially near the wavelength 1.3 μm and/or 1.5 μm, optical fiber communications can adopt various communication techniques such as space division, time division and wavelength division multiplexing techniques, which are thought to be key techniques for the realization of broadband integrated service digital networks. Integrated waveguide optical switches are thought to be key elements for the development of such technology. There have been considerable efforts to make such integrated waveguide optical switches on LiNbO₃ substrate 11-31. Though LiNbO₃ waveguide optical switches have some atractive features such as low propagation loss and low switching power because of its high electrooptic coefficient. there is no possibility of its integration with high speed electronic circuits and active devices such as laser diode, photodetector, and optical amplifier which are made on III-V compound semiconductors such as GaAs and InP. Therefore much efforts have been concentrated on the fabrication of the III-V semiconductor waveguide [4] related optical switches | 5-10 |.

There are several kinds of semiconductor optical switches suitable for a matrix optical switch, such as a carrier injection type 15,61 , a total internal reflection type 171 , a gain guided type 181 , and a directional coupler type $^{19-111}$. A reversed $\Delta\beta$ type electrooptic directional coupler(EODC) is thought be suited for a unit device of matrix switch, because its high speed operation, low electric power consumption, and little independence of input carrier wavelength. The characteristics of an

EODC is much dependent on the variations due to device fabrication [10]. Since the epitaxial layer growth techniques of GaAs/AlGaAs semiconductor and the GaAs-based device fabrication techniques are mature, it is better to choose the GaAs-based epitaxial structure as the substrate for an EODC matrix optical switch.

To get an optimum design for a GaAs/ AlGaAs 4x4 matrix switch in the viewpoint of optical loss, it is desired to consider the loss due to a doping concentration of each epitaxial layer, waveguide bending loss, and scattering loss. Because the scattering loss is dependent on the fabrication processes, it is difficult to consider the scattering loss in the design of the switch. The waveguide bending loss is dependent on the dimensions of waveguide and bending radius. Moreover, the switching voltage of an EODC is a function of waveguide dimensions [10]. Therefore it is desirable to analyse the waveguide including loss. A transfer matrix method(TMM) [11] is used to analyse the GaAs/AlGaAs multiple waveguide. This method can treat not only the multiple layer structure including metal electrode but also the loss due to its doping concentration.

There are many structural parameters to be considered in the design of the 4x4 matrix optical switch such as waveguide parameters, EODC's parameters, and bending parameters. They are closely related to each other. One have to consider all these parameters to design a low loss 4x4 matrix optical switch. A low loss substrate design is essential to accomplish this purpose. This is illustrated in section II. In section III, bending loss dependence on the structural parameters such waveguide parameters and bending parameters, is described. A cross-point device, reversed AB EODC switch determines many of the chracteristics of matrix optical switches. In section IV. The chracteristics of this type of switch, such as switching voltage and device length, and their dependences on the structural parameters of EODC are described. Using the analysis of its loss dependence on the structural parameters, an optimal design of the 4x4 matrix switch is dicussed in section V. Summary is presented in section VI.

II. Substrate design

It is desired for the device to have a low operating voltage. For this purpose, the guiding region of the waveguide should be thin. In this configuration, a pin type substrate structure is favorable because, for a constant external bias voltage, a higher electric field in the intrinsic layer can be obtained by reducing the thickness of the intrinsic layer. The switching voltage of the EODC switch is closely related to the overlap between the electric field and the optical field. The optical loss is also a function of confinement factor. Therefore it is important that the optical wave should be highly confined in the intrinsic region of the pin type substrate.

The real part of the dielectric constant of a III-V compound semiconductor is given [13] as

$$\epsilon_1 = A \Big\{ f(\chi) + \frac{1}{2} [E_0/(E_0 + \Delta_0)]^{-3/2} f(\chi_{so}) \Big\} + B \qquad (1)$$

$$\chi = \hbar \omega / E_0$$

$$\chi_{so} = \hbar \omega / (E_0 + \Delta_0)$$

$$\chi_{so} = \pi \omega / (E_0 + \Delta_0)$$

,where ω is angular frequency of the optical wave, and other parameters given for $Al_xGa_{1-x}As$ semiconductor, as a function of Al mole fraction, x, are as follows.

$$E_0(x) = 1.425 + 1.155x + 0.37x^2 \tag{2}$$

$$\Delta_0 = 1.765 + 1.115x - 0.37^2 \tag{3}$$

$$A(x) = 6.3 + 19.0x \tag{4}$$

$$B(x) = 9.4 - 10.2x \tag{5}$$

The refractive index of $Al_xGa_{1-x}As$ can be expressed as

$$n(x,\omega) = \epsilon_{1(x,\omega)}^{1/2}. \tag{6}$$

The maximum thickness of the GaAs guiding layer for the fundamental mode can be calculated for the 3 layer waveguide whose cladding layer is $Al_xGa_{1-x}As^{+14+}$. Fig. 1 shows the cutoff thickness of GaAs guiding layer as a function of the Al mole fraction x for a TEO mode of 1.3 μ m wavelength. For the doped semiconductor, there is a reduction in its refractive index and increase in its absorption. Such changes have been explained by free carrier effect [14], and expressed as

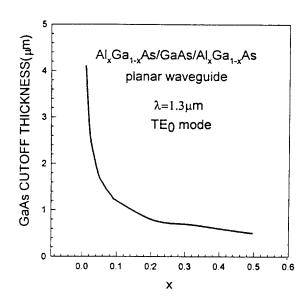


그림 1. Al_xGa_{1-x}As/GaAs/Al_xGa_{1-x}As 평판 도파 로에서 1.3 μm 파장의 TE₀ 모드를 허용하 는 GaAs 도파층의 Al mole fraction에 따른 허용두께

Fig. 1. Cutoff thickness of GaAs guiding layer of $Al_xGa_{1-x}As/GaAs/Al_xGa_{1-x}As$ planar waveguide for 1.3 μ m wavelength TE_0 mode.

$$\Delta n = -\frac{Ne^2}{2n_0\epsilon_0 m^* \omega^2} \tag{7}$$

and
$$\alpha = \frac{Ne^3\lambda^3}{4\pi n_0 m^2 u \epsilon_0 c_0^3}$$
 (8)

, where Δn is index change, α is absorption coefficient. N is doping concentration, e is electronic charge, n_0 is refractive index, ε_0 is permittivity of vacuum, m^* is effective mass of free carrier, μ is mobility of free carrier, and c_0 is speed of light in vacuum. The complex refractive index is expressed as

$$n = n_r + in_i \tag{9}$$

where n_r is the real part of the refractive index and n_i is the imaginary part of refractive index which is related to the absorption and/or the gain in the waveguide. For the lossy medium, n_i is termed as extinction coefficient and is a negative value. The extinction coefficient is expressed in terms of absorption coefficient as

$$\alpha = -\frac{4\pi n_i}{\lambda} \tag{10}$$

, where λ is optical wavelength. Since the refractive index and absorption coefficient(or extinction coefficient) can be calculated in the transparent wavelength region with the above equations, the complex refractive index of each epitaxial layer of a multiple layer waveguide can be obtained. After the calculation of complex refractive index of each layer, the effective refractive index of a GaAs/AlGaAs multiple layer waveguide can be obtained as a complex number through the TMM^[11]. Since the imaginary part effective refractive index is related absorption coefficient as Equation (10), one can obtain a propagation loss for certain multiple layer waveguides.

In the viewpoint of the coupling between the waveguide and the optical fiber, thinner waveguide may cause higher coupling loss because of mode mismatch [15]. The fiber coupling loss is mainly dependent on the waveguide dimension. Though a wider and thicker waveguide reduces the coupling loss, it requires an EODC switch with a higher operating voltage [10]. The epitaxial layer structures of these two substrates are shown in Fig. 2. As shown in Fig. 2, this structure has intrinsic AlGaAs layers to minimize the optical loss due to its free carriers.

0.2
$$\mu$$
m p⁺- GaAs (P⁺=2E18/cm)
0.6 μ m p- Al_{0.5} Ga_{0.5} As (p=3E17/cm³)
0.3 μ m i- Al_{0.5}Ga_{0.5} As
0.3 μ m i- GaAs
0.2 μ m i- Al_{0.5}Ga_{0.5} As
1.5 μ m n- Al_{0.5} Ga_{0.5} As (n=5E17/cm³)
n⁺- GaAs (n⁺=2E18/cm)

그림 2. 기판의 epi 구조

Fig. 2. Epi-layer structure of substrate.

The effective complex refractive index of structure is calculated with the TMM 111 at 1.3 µm wavelength. The substrate structure permits only a fundamental mode. Fig. 3 shows the mode intensity distribution, which exhibits a single mode characteristics. Propagation loss of 0.537 dB/cm is calculated. The fiber coupling loss is mainly dependent on the mode mismatch between the optical fiber and the channel waveguide [15]. The core diameter of a single mode optical fiber is about 9 µm at 1.3 µm wavelength. Though the maximum lateral size of semiconductor rib waveguide for a fundamental mode is a function of rib-height and its vertical structure, it is possible to choose the lateral size as 2.5 µm with a moderate manner. With these parameters including the vertical structures of the two types of substrate, coupling loss of 11.98 dB is calculated. Though this structure has relatively high coupling loss, it would have small switching voltage because of its thin intrinsic layer.

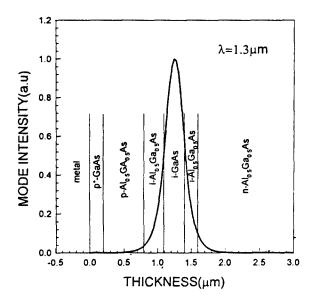


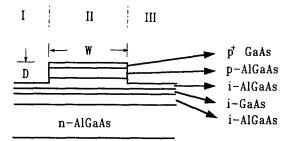
그림 3. 기판에서 1.3 µm 파장의 TE₀ 모드 도파 광에 대한 도파모드 세기분포

Fig. 3. Mode intensity distribution for 1.3 μm wavelength TE₀ mode.

II. Design of rib waveguide and bending structure

Fig. 4 shows the schematic diagram of a rib waveguide. The rib-height, D is controlled by an etching process. Since the effective refractive indices of region I, II, and III can be calculated with the TMM as described before, the effective complex refractive index of rib waveguide can also be calculated with the TMM.

There is radiation loss in a waveguide bending structure. Such a radiation loss is a function of the bending structure, and the waveguide parameters. The attenuation coefficient of the radiation loss due to its S-shaped bending structure is given [16] by



n[†] GaAs substrate

그림 4. Rib 광도파로의 개략도

Fig. 4. Schematic diagram of rib waveguide.

$$\alpha_{r} = \frac{\frac{1}{2q} \cos^{2}(\rho W/2) \exp(-2q \frac{\beta_{H} - \beta_{I}}{\beta_{I}} R) 2\lambda \exp(qW)}{\left\{ \frac{W}{2} + \frac{1}{2\rho} \sin(\rho W) + \frac{1}{q} \cos^{2}(\rho W/2) \right\} W^{2}}$$
(11)

$$q^2 = (N_{eff}^2 - n_I^2) k_0^2$$
 (12)

$$p^2 = (n_H^2 - N_{eff}^2) k_0^2$$
(13)

$$\beta_I = n_I k_0 \tag{14}$$

$$\beta_{II} = N_{eff} k_0 \tag{15}$$

where k_0 is wavenumber in vacuum, W is waveguide width. R is bending radius. β_I and β_I are propagation constants in region I and in region II, respectively. As seen in Equation (11), the bending loss depends exponentally on the bending radius and on the waveguide parameters such as waveguide width, rib-height, and its vertical structure.

Fig. 5 shows a schematic diagram of the S-shaped bending structure. After a simple calculation, one can obtain the total path length in the curvature, L_a and length L as a function of bending parameters, when bend offset T is smaller than bending radius R, as

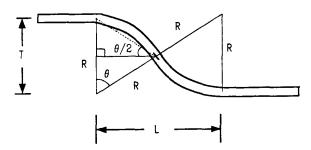


그림 5. S자형 굴곡구조의 개략도

Fig. 5. Schematic diagram of S-shaped bending structure.

$$L_{ct} = 4R\sin^{-1}\left(\frac{1}{2}\sqrt{\frac{T}{R}}\right) \tag{16}$$

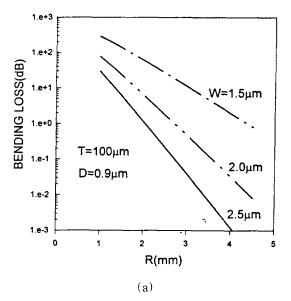
$$L = 2\sqrt{R^2 - (R - T/2)^2} \tag{17}$$

Since the waveguide parameters should be determined in accordance with the characteristics of the EODC switch, the bending loss is mainly dependent on the bending radius. For the constant waveguide parameters, bending loss can be calculated.

Fig. 6 shows the dependence of the bending loss of an S-shaped bending structure on the waveguide parameters for T=100 µm. The bending loss exponentially decreases as the bending radius increses. The bending loss decreases with the waveguide width, as seen in Fig. 6-(a). Therefore the waveguide width should be as wide as the cutoff width of the fundamental mode, in the viewpoint of the bending loss. As seen in Fig. 6-(b), the bending loss also decreases with the ribheight. These indicate, that higher confinment condition results in lower bending loss.

One may conclude that, only for the bending structure, larger bending radius, wider wavguide width, and deep etching in a rib waveguide fabrication results in low bending loss. However, this conclusion may lose its validity in the whole design of 4x4 matrix optical switch, because the

characteristics of the EODC switch, such as switching voltage and coupling length are dependent on the waveguide parameters. Therefore such parameters should be determined in accordance with the design of the EODC switch.



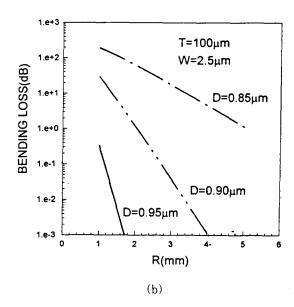


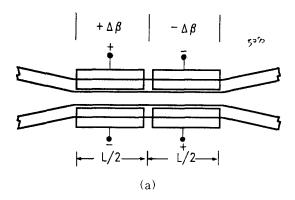
그림 6. S자형 굴곡구조의 (a) W 및 (b) D에 대한 굴곡손실

Fig. 6. Bending loss of S-shaped bending structure with respect to

(a) W and (b) D.

W. Design of reversed Δβ type EODC switch.

Fig. 7 shows the schematic diagram of reversed $\Delta\beta$ type EODC switch. The conditions for a cross(\otimes) state and for a bar(\ominus) state are expressed | 177|, respectively, as



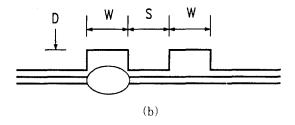


그림 7. (a) 역 AB형 EODC 스위치 및 (b) 단면 도

Fig. 7. Schematic diagram of (a) reversed Δβ type EODC switch and (b) its crossectional view.

$$\frac{x^2}{x^2 + \delta^2} \sin^2\left(\frac{L}{2}\sqrt{x^2 + \delta^2}\right) = \frac{1}{2} \tag{18}$$

$$\left(\frac{L}{l_0}\right)^2 + \left(\frac{\Delta \beta L}{\pi}\right)^2 = (4\nu)^2 \tag{19}$$

, where

$$\Delta\beta = \frac{1}{2} k_0 n^3 r \frac{V}{d} \tag{20}$$

$$l_0 = \frac{\pi}{2x} \tag{21}$$

$$L = \eta l_0. \tag{22}$$

x is coupling constant of EODC, L is device length of EODC. l_0 is coupling length of EODC.

 ν is an integer, $\delta(=\Delta\beta/2)$ is mismatch. r is electrooptic coefficient (=1.6x10-12m/V^[18]) of GaAs and/or AlGaAs, n is refractive index of GaAs and/or AlGaAs, V is applied voltage, and V is applied voltage, and V is applied voltage. By assuming that the overlap integral between light field and applied electric field is a unit. v state switching voltage can be calculated, for v = 1 and v =

$$V_{\hat{\neg}} = \frac{4\sqrt{16 - \eta^2 x d}}{\eta n^3 r k_0} \,. \tag{23}$$

For $\delta^2 = \eta^2 x^2$, Equation (18) can be expressed as

$$\frac{1}{1+\xi^2} \sin^2\left(\frac{\eta\pi}{4}\sqrt{1+\xi^2}\right) = \frac{1}{2}. \tag{24}$$

Therefore ⊗ state voltage is expressed as

$$V_{\otimes} = \frac{4\xi xd}{n^3 rk_0} \tag{25}$$

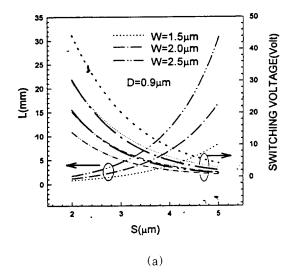
, when ξ satisfies Equation (24). One can calculate ξ numerically as a function of η . Since the coupling constant x, is given [14] as

$$x = \frac{2q^2p\exp(-pS)}{\beta(q^2 + p^2)(W + 2/p)}$$
 (26)

, one can calculate the coupling length. bar state voltage and cross state voltage as a function of the geometrical parameters of the EODC.

The switching voltage and the device length of a reversed $\Delta\beta$ type EODC switch are shown in Fig. 8. The switching voltage decreases exponentially with the separation between the waveguides. Since the device length of a reversed $\Delta\beta$ type EODC switch is inversely proportional to switching voltages, as seen in Equation (21), (22), (23), and (25), it exponentially increases with the separation of the waveguides. Since the switching voltage is proportional to η , it increase with η and the device length decreases. One can see other

structural dependences of the switching voltage and device length in Fig. 8-(a) and (b). The switching voltage decreases with the waveguide width and the rib-height.



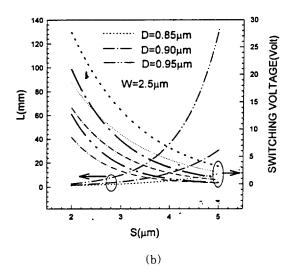


그림 8. 역 Aβ형 EODC 스위치의 (a) W 및 (b) D에 대한 스위칭 전압 및 소자 길이. 굵은 선은 ⊗ 상태의 스위칭전압을 가는 선은 ⊖ 상태의 스위칭전압을 나타낸다.

Fig. 8. The switching voltage and the device length of reversed $\Delta\beta$ type EODC switch with respect to (a) W and (b) D. Thick lines represent \otimes state voltages and thin lines represent \oplus state voltages.

The device length increases with the waveguide width and the rib-height.

Since the bending loss decreases with the waveguide width and the rib-height, one may conclude that wider waveguide width and larger rib-height result in lower loss in a 4x4 matrix optical switch. However, since wider waveguide width and larger rib-height require longer device length, these conditions result Since the in higher propagation loss. breakdown voltage of the substrate and a real propagation loss including scattering loss depend on substrate characteristics and device fabrication techniques, it might be feasible to determine the characteristics of the reversed Δ β EODC switch such as switching voltage. which can be determined by the structural parameters. After determining the structural find optimum parameters. one can an condition by reducing the losses due to passive parts of 4x4 matrix optical switch. These losses are closely related to the architecture of the 4x4 matrix optical switch.

V. Design of 4x4 matrix optical switch

Since 4x4 matrix optical switches have cross-point devices and bending many structures to connect cross-point devices, its architecture affects the loss and the cross talk between the output ports. There are several kinds of matrix switch architecture such as cross bar, tree, and simplified tree architecture [19.20]. As pointed out by K. Komatsu et al. [11], the simplified architecture have adventages of smaller path dependence in propagation loss, a small number of unit cross devices and bending structures, and low cross talk level. Therefore it is feasible to adopt the simplified tree architecture as an architecture for the 4x4

matrix optical switch.

Fig. 9 shows the 4x4 matrix optical switch whose architecture is the simplified tree architecture. This architecture has 12 EODCs and numerous waveguide bends. As illustrated above, one can calculate all the bending parameters such as the optical path length in bending structure and its longitudinal length, as a function of its bending radius, R and its bend offset. T. Since the smaller bend offset results in lower bending loss, small bend offset is feasible to design low loss matrix optical switch. In practice, considering the coupling between the waveguide and the optical fiber which is mounted on V-grooved Si substrate, one can get the smallest bend offset of 100µm^{21} to maintain the length between the adjacent ports as same.

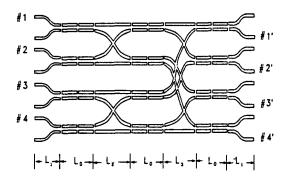


그림 9. 4x4 매트릭스 광스위치의 단순 트리 구 조의 개략도

Fig. 9. A simplified tree architecture of 4x4 matrix optical switch.

As seen in Fig. 9, the 4x4 matrix optical switch has 4 kinds of waveguide bend, whose bend offsets are different from each other. If one makes the radii of these waveguide bends large in order to reduce this bending loss, the total loss will increase because the larger bending radius cause the longer optical path length and results in higher propagation loss.

Though these different bend offsets have different bending losses, the differences of bending loss are negligible for large R. Therefore it is feasible to adopt equal R for these waveguide bends.

The total device length is expressed as,

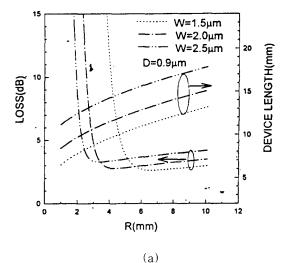
$$L_{tot} = 3L_0 + 2L_1 + L_2 + L_3 \tag{27}$$

, where L_0 is a device length of EODC switch. L_1 , L_2 , and L_3 can be calculated by Equation (17). The optical path length for each bend can also be calculated by Equation (16). The total device length and the total optical path length are functions of R. Therefore, it is simple to calculate the total loss for each optical path, including propagation loss and bending loss as a function of R. The optical path from input port #1 to output port #4' results in the maximum loss for the constant device parameters including waveguide parameters and bending radius. One should minimize this loss in order to design low loss 4x4 matrix optical switch.

The total loss and the total device length are shown in Fig. 10 as a function of R for $T=100~\mu m$. It is noted that there is an optimum bending radius, R for the constant waveguide parameters and the sepearation between the waveguides of EODC. This optimum R is determined by the waveguide parameters such as W and D.

For smaller radius than the optimum R, the total loss depends mainly on the bending loss. However, for larger radius than the optimum R, since the total loss depend mainly on the propagation loss, the increasing ratios of the total loss is equal for all data in Fig. 10. As for a loss difference between the loss due to the optical path from #1 port to #4' port and the loss due to the optical path from #1 port to #1' port, this loss difference is at most 0.1 dB for larger radius than the optimum R.

Though the propagation loss of the substrate used in this calculation is $0.537~\mathrm{dB/cm}$, there must be a additional loss mainly due to scattering loss. This additional loss is smaller than $1~\mathrm{dB/cm}^{1221}$. Even considering an $1~\mathrm{dB/cm}$ additional loss, the optimum R value is not changed.



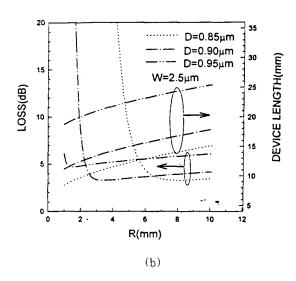


그림 10. T=100 μm, η=1.5 및 S=2.5 μm 일 때 (a) W 및 (b) D에 대한 총 손실 및 4x4 매트릭스 광스위치의 소자길이

Fig. 10. The total loss and the device length of 4x4 matrix optical switch with respect to (a) W and (b) D for $T=100 \ \mu m$, $\eta=1.5$, and $S=2.5 \ \mu m$.

Therefore one can find this optimum bend radius in order to design low loss matrix optical switch regardless of additional loss.

It is recommendable to design a low loss matrix optical switch that the waveguide parameters should be determined accordance with the requirement of the switching voltage of the reversed AB EODC switch, within the permission of a substrate breakdown voltage. After the design of the reversed AB EODC switch, an optimum bending radius can be found easily. One should also consider the tolerences which should be accompanied during the device process such as lithography and etching process because these tolences may deteriorate the performances of matrix optical switch.

If these tolerences make the waveguide width and the rib-height larger than their designed value, there will be a little increase in the total loss of the matrix optical switch for the optimum R determined by the designed value, as seen in Fig. 10-(a) and (b). However. in practice, there possibilities for waveguide parameters to be smaller than their designed value, especially in the process of photolithograpy for a long device. In this case, there must be a considerable increasing of the total loss. Therefore it is recommendable, to reduce such possiblities of drastic increasing of the total loss, that the waveguide parameters such as waveguide width W and rib-height D should be as large as the value permitted by the fundamental mode cutoff condition. It is also recommendable that bending radius R should be larger than the optimum value which is determined by the waveguide parameters. With such design rules, a matrix optical switch which has low switching voltage and low total loss can be made with a process tolerent manner.

According to the above design rule, one can design a waveguide whose waveguide width W and rib-height D are 2.5 μ m and 0.9 μ m, respectively. The separation between the waveguides of EODC, S can be chosen to be a 2.5 μ m, which results in V \otimes =8.6 volts and V \oplus =12.9 volts for η =1.5. In this case, the total loss is about 3 dB and the device length is about 14 mm for bend offset T=100 μ m as seen in Fig. 11-(c) and (d).

VI. Summary

A loss pin type GaAs/AlGaAs heterostructure substrate is designed for a 4x4 matrix optical switch. The analsis of this structure is carried out with the TMM. This substrate has a 0.537 dB/cm loss at $1.3~\mu m$ wavelength TEO mode light when considering only a free carrier effect.

The relation between the parameters of a S-shaped rib waveguide bending structure such as bending radius, its bend offset and its longitudinal length is presented. The waveguide bending loss is calculated as functions of bending radius and its waveguide parameters such as waveguide width and rib height, in conjunction with the analysis of a rib waveguide through the TMM. The bending loss decreases exponentially with the bending radius. The bending loss also decreases for higher confinement conditions such as wide waveguide width and large rib-height. For the constant bending radius and waveguide parameters, the bending loss increases with bend offset because of its optical path length.

A reversed Δβ type EODC switch is adopted as a cross-point device for the 4x4 matrix optical switch. The analysis of this switch is carried out by a coupled mode theory in conjunction with the TMM. A coupling length and a switching voltage are calculated as

functions of waveguide parameters. separation between the waveguides of EODC. and n. The coupling length of EODC increases exponentially with the separation between the waveguides of EODC. The coupling length also increases with higher confinement conditions such as wider waveguide width and larger rib-height. The switching voltage of the reversed Δβ type EODC switch is decreases exponentially with the separation of waveguides of EODC. The switching voltage decreases with higher confinement conditions. The smaller length parameter n results in the higher switching voltage because of its shorter device length.

A simplified tree architecture is adopted for a architecture of the 4x4 matrix switch. The loss and the device length of the 4x4 martrix optical switch are calculated as functions of waveguide parameters and bending parameters. There is an optimum waveguide bending radius for some waveguide parameters and EODC's parameters. Therefore one should find an optimum bending radius to design a low loss 4x4 matrix optical switch when the switching voltage of cross-point device, the reversed AB EODC switch is determined by waveguide parameters in accordance with the substrate breakdown voltage. Considering the process tolerences, it is recommended that higher optical confinement conditions such as wide waveguide width and higher rib-height should chosen for a low loss and a process insensitive 4x4 matrix optical switch. With such a design rule, a 14-mm long 4x4 matrix optical switch which has a 3 dB loss and a 12 volt operating voltage is designed.

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