

# Electrostatic Discharge (ESD) and Failure Analysis: Models, Methodologies and Mechanisms for CMOS, Silicon On Insulator and Silicon Germanium Technologies

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**Abstract**— Failure analysis is fundamental to the design and development methodology of electrostatic discharge (ESD) devices and ESD robust circuits. The role of failure analysis (FA) in the models, methodology, band mechanisms evaluation for improving ESD robustness of semiconductor products in CMOS, silicon-on-insulator (SOI) and silicon germanium (SiGe) technologies will be reviewed.

**Index Terms**— Reliability, Electrostatic, Discharge, CMOS, SOI, SiGe

## I. INTRODUCTION

Failure analysis is invaluable in the learning process of ESD protection design and development. Today, there is still no design tool which will predict ESD protection levels in a semiconductor chip for human body model (HBM), machine model (MM) and charged device model (CDM) simulation tests. ESD prediction is a difficult task because ESD phenomena spans both on microscopic and macroscopic physical scale. ESD phenomena involves semiconductor device, circuit, and package effects and their interactions. Although

significant resources have been placed on semiconductor design tools, ESD analysis and prediction remains significantly behind other circuit tool development. As a result, failure analysis assist the design and development process by providing visualization of the mechanisms leading to ESD failure. In this paper, we will focus on models, methodologies and mechanisms associated with ESD failures and failure analysis by use of examples from semiconductor technologies. The scope of the paper will show examples of failure analysis and characterization of CMOS, silicon-on-insulator (SOI) and silicon germanium (SiGe) technologies. In our discussion, we will present how failure analysis can assist in providing insight into achieving ESD robust implementations.

## 2. FAILURE MODELS

ESD failure analytical models were established to provide predictive capability for semiconductor devices. ESD failure models typically are based on the relationship of the thermal physics and pulse parameters [1-5] or statistical analysis [6-9]. ESD physical models of devices explored the relationship between the ESD pulse and the thermal transport and its role in prediction in the power-to-failure. Wunsch focused on the time scale where the thermal diffusion time is on the order of the pulse width demonstrating a  $1/t_r^{1/2}$  dependence [1]. Tasca analyzed the power-to-failure,  $P_i$ , in a spherical source. Tasca explored the dependence on the power-to-failure when the pulse is significantly smaller than the

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pulse width, demonstrating an adiabatic dependence with a  $1/t_f$  form [2]. Tasca also showed that as the pulse width is much greater than the thermal diffusion time, the power-to-failure is time independent. Today, the work of Tasca and Wunsch-Bell together are integrated into a single model to analyze the power-to-failure over all time regimes. Arkhipov, Astvaturyan, Godovsyn, and Rudenko developed a model assuming the discharge follows a cylindrical discharge phenomenon [4].

$$P_f = \frac{4\pi Ka(T_c - T_o)}{\left[ \ln\left(\frac{t_f}{b^2/4\pi D}\right) + \ln\left(\frac{4}{\pi}\right) \right]}$$

Dwyer also extended the Wunsch-Bell model by addressing the three dimensional nature of the thermal transport as three different thermal diffusion time scales  $t_a$ ,  $t_b$ , and  $t_c$  [5]. The work of Arkhipov et al influenced the direction of the Dwyer model by the introduction of the logarithm term.

$$P_f = \frac{MC\rho(T_c - T_o)}{t_f}$$

$$P_f = \frac{ab\sqrt{(\pi K\rho Cp)(T_c - T_o)}}{\sqrt{t_f} - \sqrt{t_c}/2}$$

$$P_f = \frac{4\pi Ka(T_c - T_o)}{\ln(t_f/t_b) + 2 - (c/b)}$$

$$P_f = \frac{2\pi Ka(T_c - T_o)}{\ln\left(\frac{a}{b}\right) + 2 - \frac{c}{2b} - \sqrt{\frac{t_a}{t_f}}}$$

Another method of prediction is from a statistical approach. For ESD field failure prediction, the power-to-failure distribution can be compared to the actual pulse power distribution [9]. Assuming a Gaussian distribution of the design parameters, a probability distribution function can be defined for the power-to-failure.

$$f_{P_f}(P_f) = \frac{1}{\sqrt{2\pi} S_p} \exp\left\{-1/2\left[\frac{P_f - \langle P_f \rangle}{S_p}\right]^2\right\}$$

where the power-to-failure,  $P_f$ , is the random variable

and mean power-to-failure  $\langle P_f \rangle$ , and standard deviation,  $S_p$ . The standard deviation can be expressed as

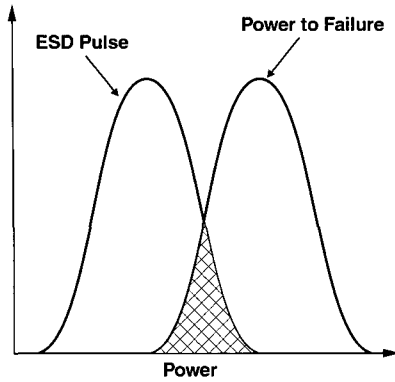
$$S_p = \left[ \frac{1}{N-1} \sum_{i=1}^N \{P_{f_i} - \langle P_f \rangle\}^2 \right]^{1/2}$$

The net field failure is the cumulative distribution function of the probability of the pulse power is greater than the power-to-failure of the device. Models have been developed by Alexander [7], Enlow [8], and Pierce-Mason [9] to provide predictive ESD results in semiconductor chips. Pierce and Mason applied this concept by assuming a Wunsch-Bell time dependence as the probability distribution and addressed dimensional variation as cause of the statistical variation.

Although these analytical and statistical models exist in the industry, a significant amount of ESD analysis today relies on failure analysis. The existence of the models, both analytical and statistical, has not assisted in the prediction and assurance needed to achieve ESD results. As a result, failure analysis plays a fundamental role in achieving successful ESD results. Failure analysis combined with other methodologies and techniques is key to the ESD verification and learning process.

### 3. FAILURE ANALYSIS AND HEURISTIC DESIGN METHODOLOGY USING DESIGN SYMMETRY

A goal of designing ESD protection networks and ESD-robust circuits is to provide uniform current distribution throughout a device. This can be achieved using three dimensional (3-D) electro-thermal device simulation [10]. Although 2-D and 3-D electro-thermal semiconductor device simulators exist, today, most ESD networks are still developed experimentally. The primary reason is the availability, development time, calibration effort and accuracy of simulation tools. As a result, experimental design and failure analysis plays a key role in the ESD design process today. As a methodology of developing ESD protection networks, it is superior to use intuition, design symmetry, and failure analysis as a means of development of a good ESD protection network. Our intuition teaches us, as a first step, that



**Fig. 1.** Probability distribution functions of the ESD pulse power and the power-to-failure of a component. Integration over the cross-hatched area provides the total failure rate.

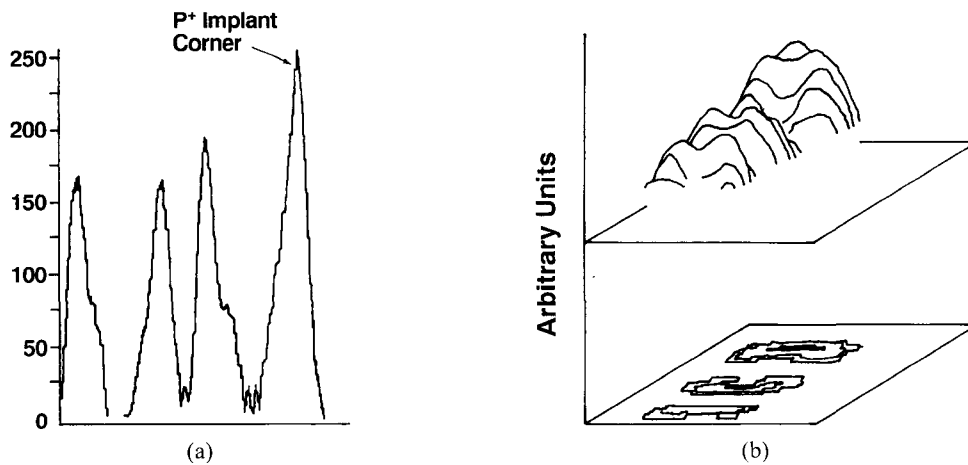
current uniformity will exist when design symmetry is self-evident in a design. Hence by providing a design whose physical layout embodies design symmetry on all design levels, it would make sense that current uniformity will occur. Failure analysis plays a key role in verifying the intuition of the ESD engineer in this first-cut assumption. In reality, there are many reasons this heuristic method is flawed. First, the actual processed on-wafer does not always represent the design level dimensions and process-induced asymmetry can be established that influence the ESD results. Second, it is hard to anticipate three-dimensional electro-thermal effects. Third, while intuition helps us with visualizing the physical layout symmetry, the “physical layout symmetry” may not be equivalent to the “electrical symmetry” or “electro-thermal symmetry”. Failure

analysis plays a key role in correcting intuition and provides bridging from the layout to the understanding of the electrical and thermal response of a device or circuit network.

The FA damage pattern itself and its distribution throughout the ESD network can be used as a “efficiency factor” or ESD metric. In actuality, a device which has a significant percentage of the area as damaged, demonstrates the ability to utilize the maximum percentage of the silicon area for providing ESD protection. The damage pattern symmetry also is an indicator of the ability to achieve current uniformity and a measure of good ESD design practices.

#### 4. FAILURE ANALYSIS METHODOLOGY AND ELECTROTHERMAL SIMULATION

Using a heuristic understanding of the design layout, failure analysis followed by electro-thermal simulation can provide a higher intuition allowing the ESD designer bridge from the physical to the electro-thermal results. Failure patterns can teach the regions of peak thermal heating and failure [10, 11]. Electro-thermal device simulation can help understand the location and the root cause of the ESD failure. As an example, the corner of an shallow trench isolation (STI) bound p+ diffusion/n-well diode was failing on the diffusion-to-STI corner. Using a scanning electron microscope (SEM), an emission microscope (EMMI) tool (Figure 2), and a



**Fig. 2.** Emission microscope (EMMI) tool photon map of a p+ diode structure pre-ESD stress.

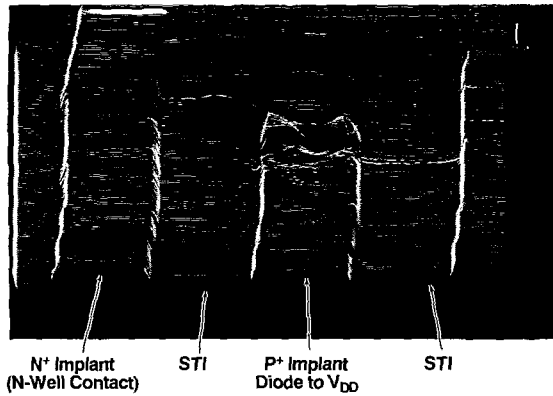


Fig. 3. Kelvin probe force microscope (KPFM) topography image of p+ diode post-ESD stress.

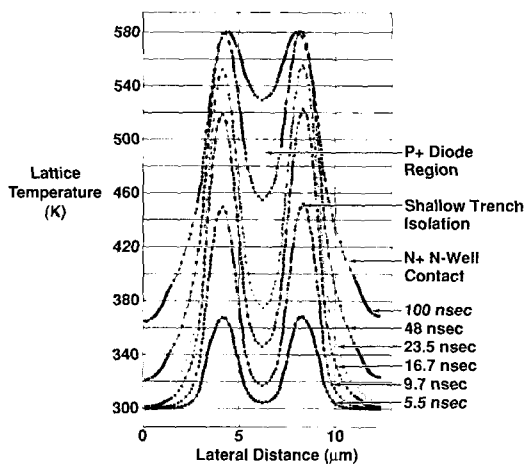


Fig. 4. Electrothermal Simulation Results.

Kelvin force probe microscope (KPFM) atomic force tool (Figure 3), the ESD damage was imaged. The SEM provided a bird's eye view of the ESD device after removal of the metal films. The EMMI tool provides a photon-mapping of the ESD structure during direct current (d.c.) measurement. The KPFM atomic force tool provided both a topographic as well as electrical potential mapping. Using a 3-D semiconductor electro-thermal tool FIELDAY III, our analysis demonstrated and verified that the peak lattice temperatures was at the end of the p+ diode implant [10].

Although 3-D electro-thermal device simulation is possible, it is extremely sensitive to the electrical and thermal Von Neumann and Dirichlet boundary conditions, mesh quality and mesh densification techniques. Even with good refinements, heat capacity and thermal conductivity models are not well

characterized or calibrated in the high temperature regimes. Independent of the calibration and model fitting, FA combined with the electro-thermal simulation establishes good intuition and a good methodology for the design of ESD protection networks and circuits.

## 5. FAILURE ANALYSIS METHODS AS A MEANS OF ESD DEVICE OPERATION VERIFICATION

Failure analysis can be used as a means of verifying ESD device operation. At times, it is not clear to all ESD engineers how an ESD device is operating or the current paths. Failure analysis is a key means of verification of the current transfer based on the location of the damage on given shapes, or between shapes. The FA damage is a verification of current transfer and clearly can show device operation and the path of current transfer. For example, in integrated cascode MOSFET, the electrical schematics would not explain the nature of the failure mechanism. Early measurements of cascode MOSFETs anticipated that the MOSFET snapback voltage would serve as the sum of the two MOSFETs. Experimental results verified that integrated series cascode MOSFET was significantly less than the sum of the two MOSFETs. It is clear from the failure analysis that the interaction for cascaded MOSFET second breakdown occurs in the same local region, providing a response which behaved as a single MOSFET. From the AFM failure analysis, it is clear that the parasitic bipolar transistor is interactive locally as one device. The AFM failure analysis results then shows that treating the series cascode MOSFET structure can not be modeled as two independent components. Since this early work, the issue of analysis of series cascode MOSFETs has had increased interest in mixed-voltage interface networks in microprocessors and peripheral circuits.

## 6. FAILURE ANALYSIS AND ESD TESTING METHODOLOGY

For failure analysis to be effective for ESD learning, a good ESD testing methodology and strategy is needed to

maximize the ESD learning on ESD networks, circuits and products. The ESD testing methodology is key in providing valuable correlation between the ESD failure and the failure mechanism. In IBM, the testing and failure analysis strategy, developed by Gross and Voldman [12], all pins are tested to a given power rail. Second, each pin is tested from zero volts on the source, and step stressed in small increments. It is also key that all pins are tested to failure. The failure distribution function of all the pins are plotted and evaluated to determine the “mean failure distribution”, standard deviation, and other statistics of the whole chip. In many corporations, the FA focus is primarily on the worst case pins. In this methodology, the failure analysis of the “good” pins are as important to evaluate as well. This allows documentation and classification of the pin types, the failure mechanism and ESD results. This methodology allows to verify at what level different failure mechanisms are evident as well as quantify what mechanisms are occurring in what circuit function. This testing methodology highly incorporates failure analysis as a key source to drive ESD learning in chip development.

## **7. FAILURE ANALYSIS AND SEMICONDUCTOR PROCESS INDUCED ESD DESIGN ASSYMETRY**

Failure analysis is a means to verify semiconductor process or design asymmetry which can have an influence on ESD robustness of circuits. Process induced design asymmetries can occur on all design levels. For example, photo and etch processing can have both macroscopic and microscopic effects [13]. In the processing of polysilicon gate structures, “across chip line-width variation,” known as ACLV, can vary the MOSFET polysilicon gate line-width. The placement of the circuit in the chip globally leads to ACLV effects because of macroscopic photo-tool source-intensity effects, and within a multi-finger MOSFET because of microscopic photo and etch effects. It is found that depending on the type of photo-resist, “nested” line-widths and “isolated” line-widths can be different. With positive tone resist, nested lines are smaller than isolated

lines, and with negative tone resist, this reverses. The implication of this effect is that MOSFET second breakdown is a non-random phenomenon and will be evident on the MOSFET fingers with smaller line-widths because of the lower MOSFET snapback. Using a two-dimensional “boot tip” atomic force microscope (AFM), two-dimensional mappings of topography provided significant accuracy of both the lateral and vertical dimensions. Boot-tip AFM devices allows for the ability to address sidewall slopes of polysilicon gate structures and vertical trench profiles. Failure analysis verified that the MOSFET second breakdown in the multi-finger MOSFET structure was non-random in a positive tone photo-resist system. Failure analysis results showed that the “nested” lines with the smaller MOSFET line-widths had damage between the source and drain. This ACLV nested-to-isolated effect was eliminated as the technology switched from positive to negative tone photo-resist [13]. Combining the 2-D AFM tool with the SEM analysis of the ESD-damaged MOSFETs, verification of the reason for the non-random component associated with line width as quantified and explain the ESD results.

## **8. FAILURE ANALYSIS METHODOLOGY TO EVALUATE INTER-POWER RAIL ELECTRICAL CONNECTIVITY**

Failure analysis can be used to verify and evaluate the electrical connectivity between two power busses and how they interact with a peripheral circuit. The ESD robustness level, response and failure analysis of a multi-finger MOSFET circuit is significantly dependent on the relationship between the chip substrate and the other ground rails. In CMOS chip, typical the off-chip driver (OCD) bus or peripheral circuits bus is separated from the core or chip substrate bus [14]. In mixed signal designs, the digital and analog grounds are also separated. In RF applications, different RF functional blocks also have separated grounds. In the case that there is strong bi-directional coupling between the grounds, the failure damage pattern in a multi-finger shows damage in many of the MOSFET fingers. When the circuit is tested referenced to a substrate-decoupled

peripheral ground, the ESD results are significantly lower and only one damage spot in a single finger is evident. Hence from the failure analysis pattern in the MOSFET, it can be determined whether is strong or weak coupling between the rails.

## 9. FAILURE ANALYSIS METHODOLOGY FOR EVALUATION OF ESD PARASITICS

Failure analysis provides significant insight into the current flow of parasitic elements and how they interact with an ESD element. Three situations can exist 1) the parasitic element draws the current, becomes damaged and leads to low failure levels, 2) the ESD network draws the current flow, becomes the limiting value and fails at a high current, and 3) the parasitic and the ESD element are working in conjunction where both are damaged at the same failure voltage. The first case is sometimes referred to as “ESD current robbing”. For example, a parasitic npn is formed between a n diffusion resistor and a guard ring structure. Additionally, an ESD n-well diode is also in parallel with the n-well diode element which is also adjacent to the same guard ring structure. In our experimental results, the parasitic npn reduced the failure level to  $-3$  kV whereas the ESD element n-well diode would have produced  $-7$  kV. At  $-3$  kV, no damage is evident in the n-well diode but in the parasitic npn damage is evident between the guard ring and the resistor element. Failure analysis is able to determine that “ESD current robbing” was occurring leading to the low failure level.

## 10. FAILURE ANALYSIS METHODS AND ESD GROUND RULES

Failure analysis is also used to develop ESD ground rules. ESD ground rules are design rules which are specifically used to minimize the effect of parasitic elements or ESD failure. Using the failure damage patterns, and ESD robustness levels, these spacings can be decided. With the example in the prior section, it was found by adjusting the spacing between the n-diffusion and the guard ring, the “ESD current robbing” could be

eliminated. It is found that an optimum ESD result is possible when the spacing of n-resistor to the n-well guard ring was tuned to the spacing between the n-well ESD diode and its adjacent guard ring. Hence an ESD ground rule was established so that the spacings provided the highest ESD level. Hence, using failure analysis damage patterns, the spacing of guard rings and ESD ground rules can be defined.

Interactions between floating-gate tie downs and the pull-down MOSFET of a mixed-voltage off-chip driver circuit also was evident using failure analysis of the peripheral circuit. The drain of the MOSFET and the n+ diffusion tie down formed a parasitic bipolar transistor when designed local to each other in the substrate. In this work, it was found that in negative VDD test modes, the lateral npn was active in LOCOS isolation, but not in STI-bound MOSFETs. Ground rules were established to avoid interaction between the floating gate tie down elements and the peripheral circuit pull-down network.

## 11. ESD FAILURE MECHANISMS, AND TECHNOLOGY EVOLUTION AND REVOLUTION

With the evolution and revolution of CMOS technology, at times some types of ESD failure mechanisms are eliminated while others are omnipresent. With technology transitions, the importance of some ESD failure mechanisms were important years ago, were resolved and again rise as a key technology concern [15-22].

In early NMOS technology, there was considerable focus on interconnects and oxides. Turner, in 1980, stated “Today’s 2 and 3 micro metal widths have reduced current carrying capacity to the point where it has become a real concern from an ESD standpoint.” In this time frame, devices were un-passivated leading to vaporization of the aluminum (Al) wires and polysilicon resistors. Aluminum and alloy spiking occurred both vertically and laterally. ESD-induced gate oxide pin holes were observed in NMOS technology as the most dominant field failure mechanism [23]. With the introduction of resistors and thick oxide MOSFETs to lower the voltage at the receiver gate oxide, these issues

were resolved.

With the introduction of silicide films, the focus shifted to MOSFET second breakdown and MOSFET ESD robustness for technologies whose ground rules are less than 2  $\mu\text{m}$ . With the introduction of silicides, the low resistance of the silicide eliminated the natural resistor ballasting in the MOSFET source and drain. With the introduction of low doped drain structures, continuous junction scaling, and silicide, the product of the current density and peak electric field lead to higher temperatures in the MOSFET drain structure. MOSFET second breakdown remained a dominant failure mechanism in technology.

For 0.5 to 0.35  $\mu\text{m}$  Leff technology, with the scaling of ground rules, significant activity was present in parasitic npn and pnp type devices. With the introduction of shallow trench isolation, ESD learning, ESD ground rules and ESD checking systems, many of the ESD concerns with these parasitics were eliminated.

With smaller design rules and increased circuit density, interconnects were scaled in width and film thickness for density and performance objectives. Interconnect scaling is driven by the MOSFET scaling dimensions. According to Rent's rule, the number of I/O increases proportional to the number of gates on the chip. As the interconnects, vias and contacts are scaled, the ESD robustness of a given interconnect decreases. For 0.25  $\mu\text{m}$  and below CMOS technologies, interconnects became the dominant failure mechanism [16,17,19]. In aluminum interconnects, the Ti/Al/Ti structure failure was typically observed as a lateral cracking of the insulator and Al filling this damaged region. Direct current (d.c.) resistance shifts occurred at 90% of the critical current. This resistance shift can lead to latent mechanisms in circuits [20]. With new structures, Banerjee constructed a model for the current-to-failure for a Ti/Al/Ti interconnect incorporating the three film structure and isolation region [18].

With the migration from Al to Copper (Cu) interconnects, the ESD robustness of interconnects improved because of Cu has a higher melting temperature compared to aluminum [19-21]. HBM, MM and TLP measurements verified the improvement in both test sites and product implementations. In a damascene process, the high refractory metal liner serves as a barrier to Cu diffusion. The failure mechanism in liner-cladded

Cu interconnects typically show cracking above the trough as opposed to lateral cracking. For interconnects below 2  $\mu\text{m}$ , lateral extrusion was observed. For dual damascene Cu interconnects, Voldman extended this model to address the three-sided trough lining of the damascene interconnect and Cu filling materials as well as the surrounding insulation [19,20,21]. For low-K materials, experimental measurements were also taken to compare SiO<sub>2</sub> and low-K material impacts on the current to failure of a damascene interconnect highlighting the ESD robustness degradation due to the low-K insulation material [21].

For CMOS technology to follow the Moore's law relationship, it has been proposed that this is achievable by having CMOS technology migrate to silicon-on-insulator (SOI) technology [24,25].

With SOI technology, the failure mechanisms are significantly different from bulk CMOS. This is a result of new ESD networks and the lack of contact to the chip substrate. With the introduction of the SOI lateral gated diode ESD element, the ESD failure was observed from the anode to the cathode under the SOI MOSFET gate dielectric. In SOI, circuit elements which normally discharged to the substrate in bulk CMOS, typically failed in the reversed mode or took alternative current paths. Unlike the bulk CMOS, no failures were observed in the inverter gate structures for the CDM mechanism. The CDM failure mechanism occurred in the gate to input node of the SOI MOSFET pass transistor. Evidence from the failure analysis showed the current path occurred through the VDD to the gate of a SOI MOSFET pass transistor as opposed to the traditional CDM failure mechanism of the MOSFET inverter gate and the pass transistor source-to-drain. New failure mechanisms similar to the pass transistor also occurred in the buried resistor (BR) structure. The BR resistor is consists of a MOSFET gate structure, where dopants exist in the channel region to form a resistor. The MOSFET gate structure serves the purpose of blocking the silicide film, isolating the resistor input and output. BR resistors which normally discharge into the substrate, can not in SOI. In SOI, these structures can fail in a negative pulse device ESD stress. CDM failure mechanisms also occur in the gate structure of the lateral Lubistor structure when the gate was connected to the

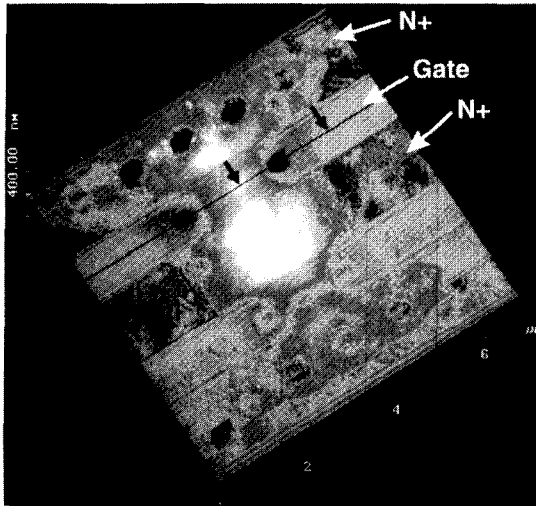


Fig. 5. MOSFET Second Breakdown in a Series Cascade MOSFET structure.

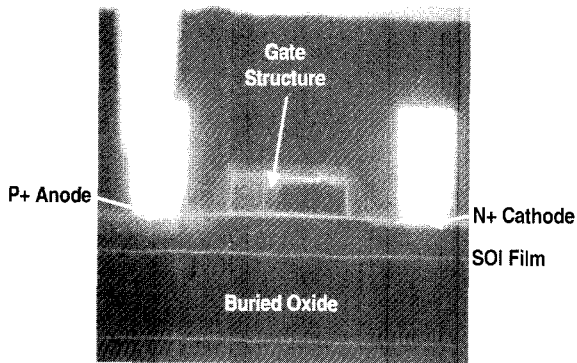


Fig. 6. Cross section of an lateral gated diode ESD structure.

cathode. Hence, isolation of the Lubistor gate structure prevents CDM failure mechanisms in these SOI structures. To resolve this issue, new SOI ESD circuit solutions were developed to eliminate the failure mechanism in the lateral gated Lubistor circuit and the BR resistor elements by electrically de-coupling these gates from the anode, cathode or power supply nodes. An alternative path, chosen by Ker, extended the ESD reliability of the lateral gated diode by removal of the gate with an additional mask [26]. Two key points to address on ESD in SOI technology. First, excellent ESD results as good as or better than bulk CMOS products is achievable in SOI semiconductor products. Second, to achieve good ESD protection in SOI, a significant amount of ESD learning is required to address the new ESD-induced failure mechanisms. These mechanisms need to be solved by either device structural, design layout or circuit modifications.

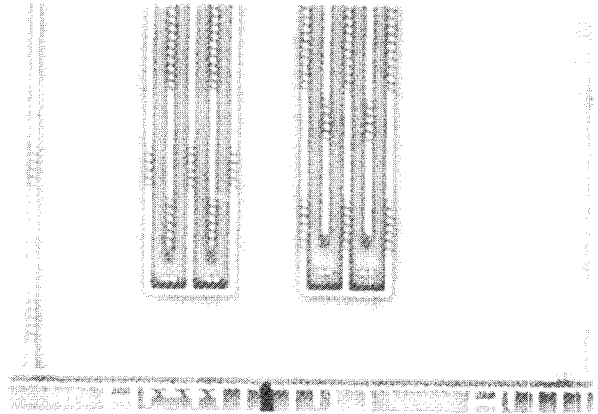


Fig. 7. Crosssection of the ESD structure after ESD failure. ESD failure is on second finger of first diode structure.

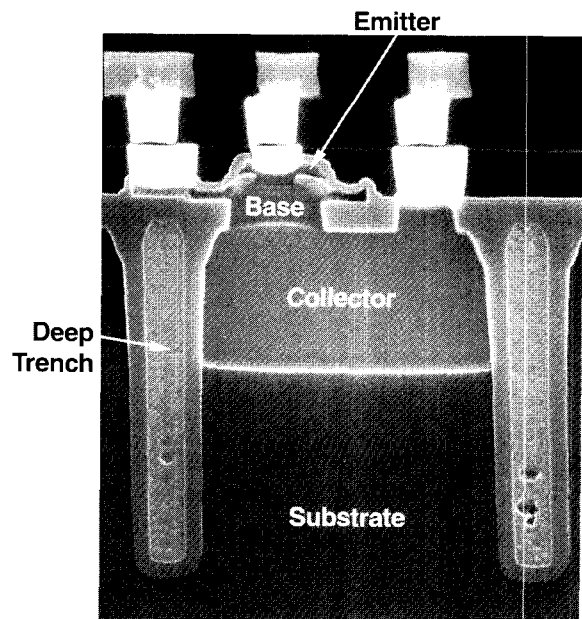


Fig. 8. Cross section of a SiGe HBT device.

## 12. RADIO FREQUENCY (RF) DEVICES AND ESD

With the growth of high speed data-rate wired communication, wireless market, high speed test equipment, and magnetic recording industry, advanced semiconductors are playing a more significant role in this millennium. RF CMOS, Gallium Arsenide (GaAs), Indium Phosphide (InP), Silicon Germanium (SiGe) and Silicon Germanium Carbon (SiGeC) technologies will dominate these application spaces of 1 to 300 GHz. Early work on ESD sensitivity of RF devices was performed by Whalen and Domingos [27,28]. ESD



sensitivity of these devices will be a larger concern as the device sizes decrease and the application frequencies increase. Although GaAs devices and GaAs technology has matured, very little improvement in the ESD protection levels have been demonstrated above 1000 V HBM.

SiGe and SiGeC BiCMOS technology has the advantage of natural integration with advanced CMOS foundry technology as well as a myriad of additional design levels, implants and devices. As a result, excellent ESD robustness levels have been demonstrated in BiCMOS SiGe technology by utilizing both CMOS and bipolar ESD elements [29-34]. CMOS ESD solutions can be incorporated in the digital, analog or RF sectors of a mixed signal RF CMOS or RF BiCMOS semiconductor chip. The CMOS ESD elements can include p+/n-well diodes, n-well-to-substrate diodes, n+ diodes, grounded gate NFET elements. BiCMOS features, such as buried layers, sub-collector and reach through implants, and trench isolation can be added to provide improved ESD and latchup robustness to these CMOS elements. Additionally, CMOS-based ESD power clamps, such as RC-triggered MOSFET circuits, can be used for ESD protection of the MS chips. To utilize these for RF CMOS or RF BiCMOS, RF models must be established for the RF and analog sectors of a SOC application.

In an RF BiCMOS product, many circuit designers prefer to use bipolar-based active and passive ESD elements because of the low noise, high Q, low capacitance, and voltage power rail considerations. Although SiGe technology research and development was started in 1980's, only recently with the increased volume of SiGe products has the interest in ESD protection has increased. Figure 9 shows the first transmission line pulse (TLP) I-V plots of a SiGe heterojunction bipolar transistor (HBT) device in a common-emitter configuration as a function of the pulse width. In the TLP measurement, the pulse magnitude was increased until SiGe HBT device failure. A set of TLP I-V characteristics were measured as a function of different pulse widths. As the pulse width is varied, it can be observed that the trigger conditions, resistances and failure current is varied. Figure 10 shows a Wunsch-Bell plot of power-to-failure versus the pulse width of an epitaxial base Si BJT and epitaxial base SiGe HBT

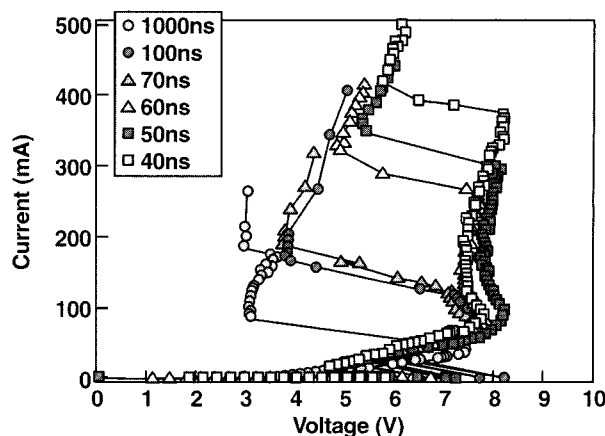


Fig. 9. TLP I-V characteristic of SiGe HBT device ( $\tau = 40\text{ns}$  to  $\tau = 1000\text{ns}$ ).

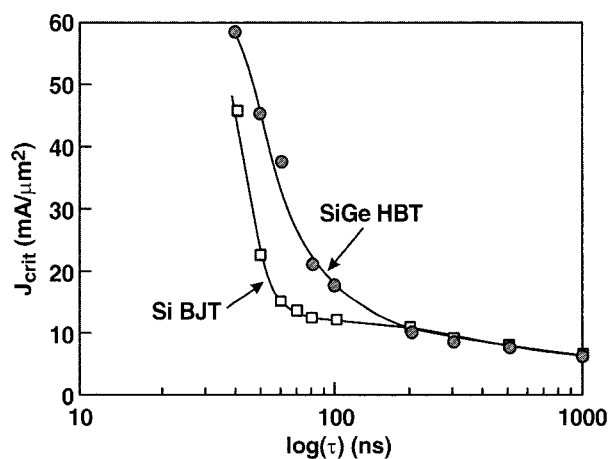


Fig. 10. Wunsch-Bell Power-to-failure plot of SiGe HBT and Si BJT devices.

device. Experimental results show that as a result of the higher current gain, the SiGe HBT device results exceed those of the Si BJT device.

The emitter-base junction of the bipolar transistor has been historically a limiting failure mechanism. Heterojunction bipolar transistors allows for an increase in the base doping concentration for a given unity current gain cutoff frequency. Figure 11 shows a TLP I-V plot and leakage history of a SiGe HBT npn in a reverse biased emitter-base operational mode. Experimental results show that as the stress exceeds the safe operating area (SOA) of a SiGe HBT device, leakage current increases and then saturates. When the breakdown voltage is reached, a rapid increase in leakage current and device failure is evident.

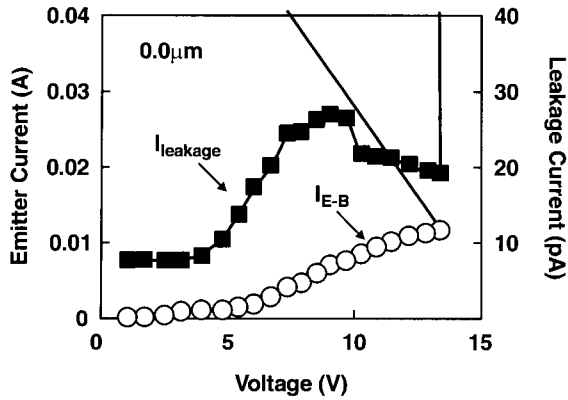


Fig. 11. TLP Measurement of NSA SiGe HBT emitter-base for large emitter diffusion-to-extrinsic base implant spacing.

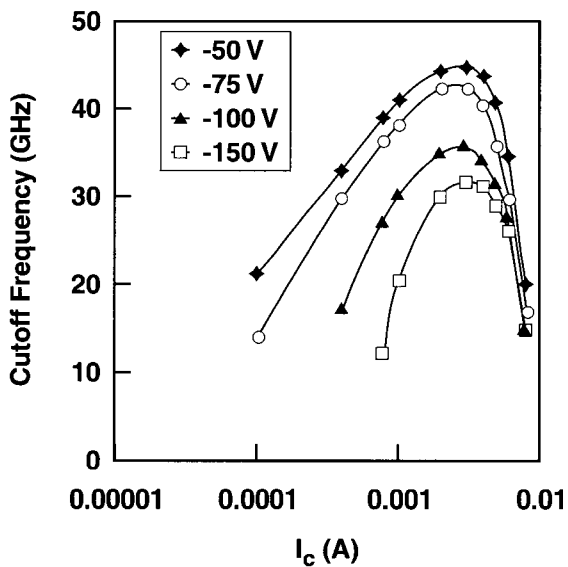


Fig. 12. Unity current gain vs collector current degradation plots with increased ESD stress in a SiGeC HBT device.

Evaluation of the RF characteristics during ESD stress is important to evaluate the onset of performance degradation. The demarcation point where RF characteristics are below semiconductor device or product specification is important in determining the failure point. In an RF product, there are a multitude of RF specifications which may be needed to determine the failure point. A study was completed where unity current gain cutoff frequency as a function of collector current ( $f_T$  vs  $I_c$ ) was evaluated prior to ESD stress on a SiGeC HBT device (Figure 12). A post-RF stress was completed to evaluate the change in the device characteristics. Experimental results infer that the transconductance is impacted by the ESD stress as well as the peak cutoff frequency. In this study, the d.c.

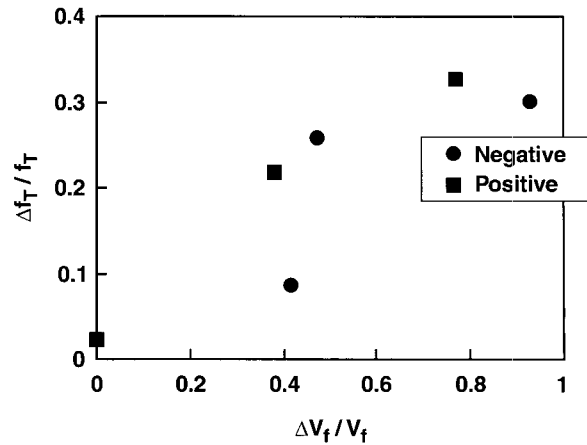


Fig. 13. Plot of the percent change in the cutoff frequency as a function percent change in the forward bias voltage.

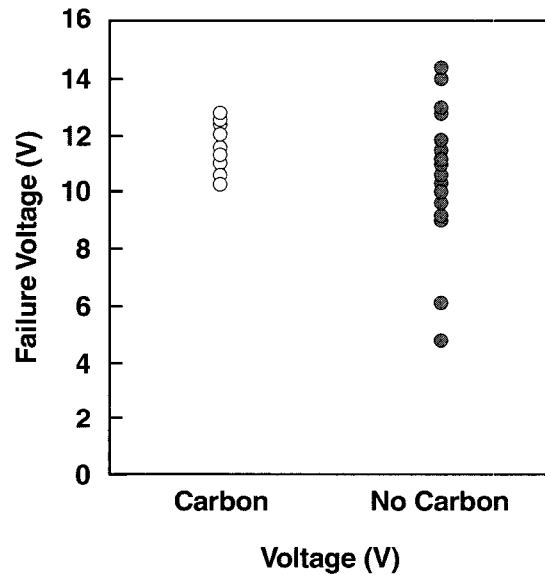


Fig. 14. Voltage to Failure of SiGe transistor with and without Carbon implant in the SiGe base region.

parameter shifted also showed degradation in the emitter-base configuration. Figure 13 demonstrates a correlation exists between the normalized shift in the dc characteristics and the cutoff frequency degradation for both positive and negative stress. A key point is that the RF degradation can be identified by the change in the dc characteristic once the correlation is established. Secondly, the failure occurs for both the positive and negative stress conditions. In RF products, there are a significant number of RF parameters and RF functional tests which will be needed to be evaluated in order to insure ESD reliability. These parameters are dependent on the product application. Power gain, gain com-

pression, output-intercept 3rd order harmonic (OIP3), linearity, quality factor and other design parameters may be needed to evaluate to guarantee ESD robustness additional to the standard d.c. shift or leakage test requirements.

In the scaling of SiGe HBT devices, base width scaling is important to achieve higher device performance. The objective is to decrease the base width and yet still provide a low base series resistance. With the incorporation of Carbon into the SiGe epitaxial film, the base doping concentration can be increased, providing improved current distribution in SiGe HBT devices and improved unity power gain cutoff frequency,  $f_{MAX}$ . Experimental results in SiGeC HBT devices and ESD power clamps have already demonstrated ESD levels above 2 kV HBM [34]. The placement of Carbon in the base region of a SiGe HBT constrains the transient enhanced diffusion (TED) of the Boron atom. As a result, resistance parameters, breakdown parameters and ESD results demonstrated tighter statistical control (Figure 14 and Figure 15).

### 13. MASKS AND ESD

ESD damage was observed in production masks due to static induction charging as early as 1983. Chemelli, Unger and Bossard noted damage patterns to production masks failure analysis pattern could be replicated by a laboratory controlled induced ESD [35]. High current at the corners lead to field emission as well as arcing across the narrow spaces.

Significant research in air ionization and installation of ionizers in mask production lines have lowered the concern for ESD-induced damage to masks in the last 17 years. With the continued scaling of mask dimensions, both permanent damage and latent mechanisms have again become a concern. A key result of recent work of Montoya et al [36] showed that when the spacing between lines are larger, the damage threshold is higher and the energy transfer is greater leading to an observable damage pattern. As the spacing decreases, the energy in the discharge decreases leading to a less observable damage pattern which may escape mask inspection -- this concern will address itself as a potential latent mask defect. New research in masks and

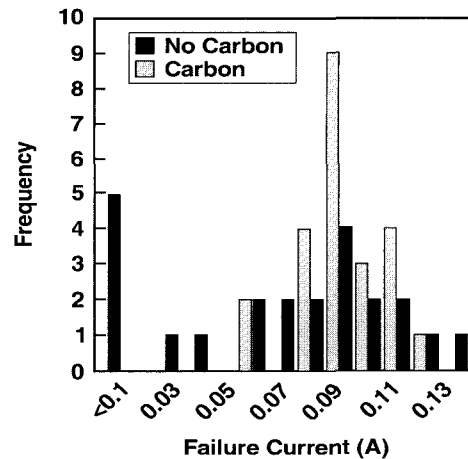


Fig. 15. Histogram of Current-to-Failure of SiGe and SiGeC HBT device.

the discharge process is addressing these concerns for future technology generations.

### 14. MAGNETIC RECORDING AND ESD

Electrostatic concerns in the magnetic recording industry has increased with the introduction and scaling of the magneto-resistive (MR) heads. In prior technology generations, a thin film inductive read/write recording head was used. With the transition to an inductive write and magneto-resistive (MR) read head, ESD concerns were evident. The publication of Wallash, Hughbanks and Voldman [37], showed several ESD failure mechanisms were evident in the MR head . A first mechanism occurs between the MR shield and the MR head element which occurs via an arc discharge across the MR-shield gap. A second mechanism occurs from the current that flows through the MR positive and negative leads. The results in this study showed the failure level decreases linearly with MR thin film stripe width. MR head failure mechanisms manifest themselves through electrical resistance increases. Additionally, in the inductive write head, failure mechanisms also show that ESD failure can occur between the pole tips and the substrate as well as internal coil and dielectric damage.

With the introduction of the Giant Magneto-resistor (GMR) head, continued scaling of the thin film stripe lowered the ESD failure voltage by 5X. The introduction of the spin valve (SV), which provides a means of programming the GMR, can also lead to a new failure

mechanism. Spin-valve reversal occurs when the ESD pulse itself influences the initial program state of the GMR head.

In future generations, new inventions, such as the Tunneling Magneto-resistive (TMR) devices also show a reduced HBM ESD failure voltage below 10 V HBM levels [38]. New challenges will be evident in the production, manufacturing and handling of these new magnetic recording devices as the ESD sensitivity decreases below the 50 V HBM levels.

## 15. CONCLUSIONS

Although technology evolution and revolution has changed, three basic mechanisms of ESD failures still exist; junction damage, dielectric damage and metallization damage. With the migration from NMOS to CMOS, from single well to twin well, from diffused wells to retrograde wells, from thick to thin oxides, from Aluminum to Copper interconnects, from high K ILD to low K ILD, from tapered Al vias to Cu dual damascene vias, from AMR to GMR to TMR - - the failure mechanisms are still similar. What does change is the relative importance of one mechanism to another; this is dependent on how devices scale, and whether the technology migration leads to an ESD improvement or ESD degradation. With the introduction of new technologies, such as silicon-on-insulator (SOI), and silicon germanium (SiGe), new failure mechanisms will occur as a result of new physical structures and new physics. With the rapid increase of interest in RF GaAs, InP, and SiGe technologies, new failure criteria for ESD failure will be to address RF functional degradation effects. With the continued scaling of semiconductors, magnetic devices, and photomasks, the new ESD issues and their failure analysis will continue to be important for device design, reliability and manufacturing of these components.

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