Active Vibration Control of a Plate Using TMS320C6713DSK

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Abstract : This paper deals with the experimental study of the vibration suppression of the smart structures. First, a new high-speed active control system is presented using the DSP320C6713 microprocessor. A peripheral system developed is composed of a data acquisition system, A/D and D/A converters, piezoelectric (PZT) actuator/sensors, and drivers using PA 95 for fast data processing. Next, the processing time of the peripheral device is tested and the corresponding test results are provided. Since fast data processing is very important in the active vibration control of the structures, achieving the fast loop times of the control system is focused. The control algorithm using PPF in addition to FIR filter is implemented. Finally, numerous experiments were carried out on the aluminum plate to validate the superior performance of the vibration control system at different control loop times. **Key words :** Active vibration control, DSP, PZT actuator/sensor, PPF control

1. Introduction

Recently, there has been widespread interest in the active vibration control system. Owing to the performance of the improved piezoelectric devices, a great deal of attention has been paid on the research and development of high-speed active control system integrated with PZT sensors/ actuators and controllers. There have been a large number of theoretical papers for vibration control of the cantilever beam and thin-plate beam. However, few papers have considered the active control system with fast processing microprocessors, experimentally.

A control system using VMEbus processor was presented for vibration control of the thin-plate [1]. Also, there has been an experimental result on real-time control for a rectangular steel plate incorporated with ADSP21062 EZKI and EZ_ANC II digital-signal-processing (DSP) [2]. In addition, Chu et. al. had designed a simulator using real-time active controller with TMS320C40 DSP board [3]. The active controller has been proposed for cylinder shell structures, integrated with PZT driver and TMS320C30 DSP using PPF controller and filtered-x LMS controller [4]. Further, the active controller for steel plate with 340x300x0.5 [mm³] in dimension has utilized DSP320C30 DSP featuring multi- adaptive feed-forward control inputs [5]. A DSP controller and selected control algorithms are implemented to а complex autoparametric "L-shape" beam system to compare numerical and experimental results. Detailed analysis for the NSC system is carried out [6]. A novel distributed actuation approach sensing and for actively

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suppressing vibrations within flexible link manipulators has been presented where UPM-2405 power amplifier, which is interfaced with a Quanser Q8 (8-channel) control board embedded in a host PC has been applied [7]. Active vibration control using modal controllers has been successfully implemented on simple structures such as beams and plates. A stand-alone DSP having sampling frequency of 20 kHz is applied and the input channels to the DSP are connected via a sensor amplifier to four PVDF film sensors [8].

In this study, to prove the performance of the fastest controller experimentally, a new active control system is developed adopting fast microprocessor in TMS320c 6713 series with peripheral devices; high-speed A/D and D/A converters, and a new PZT driver. For the control algorithm, a PPF control with FIR filter is applied.

2. Design of the control system

In PZT materials there is coupling between the electrical and mechanical constitutive relationships. Note that PZT materials generate an electric charge proportional to applied stress, which makes them a natural sensor. Conversely, PZT materials can be effectively used as an actuator, as they deform when an electrical charge is applied. As a result, they can be used in actuation and sensing.

The general constitutive equations of linear-piezoelectric materials describe a tensor relation between mechanical and electrical variables in the form [4]

$$S = s_E T + dE \tag{1}$$

$$D = dT + \varepsilon_T E \tag{2}$$

where S is the mechanical strain, T is the mechanical stress, E is the electrical field, D is the electrical displacement, S_E is the mechanical compliance of the material measured at zero electric field (E=0), ε_T is the dielectric permittivity measured at zero mechanical stress (T=0), and d is the piezoelectric coupling between the electrical and mechanical variables.

Referring to Figure 1, an active control system is composed of а high-speed plus relatively DSP320C6713 low-priced processor whose processing speed is 1500 [MPLOPS]. In other words, its processing speed is 27 times faster than the DSP320C30 processor, and its detailed described in [9]. Also, an specifications are interface system has been developed based on A/D and D/A converters providing extremely high-speed processing and resolutions. More explicitly, the complete actuator system is consisted of the PZT driver and elements: the PZT driver is intended to amplify control signals from the DSP board up to ±400 [V] for the actuating elements, and the PZT sensor detects the amplitude of vibration and transmit it to the DSP control board via the A/D converter. The MFC actuator/sensor used is NASA M8557 S1 type, and its specification is described in [10].

In this study the interface system to communicate with external devices includes A/D converter for sensing signals from PZT device and D/A converter for receiving signals from PZT actuator driver. They are ADS7805 and DAC712 chips charactering 16 bit resolutions, respectively. Figure 3 depicts the I/O interfaces ensuring communication with the main controller, A/D and D/A converters. The PZT driver amplifies control signals of the main controller to execute active control action up to PZT driving voltage. In the study, a PA97 commercial driver manufactured by APEX Corp. have been utilized as the actuator amplifier.

Now, the structure and configuration of the complete control system are shown in Figure 1 and in Figure 2. The PZT sensor is connected to

the input port of the A/D converter wired to the DSP chip whose data and address buses are further connected to data ports of the D/A Analog output the D/A converter. ports of converter are connected to the input ports of the PZT actuator amplifier, and the output ports of the non-inverted amplifier are now linked to the input ports of the PZT actuator amplifier.

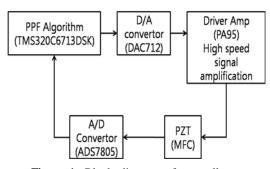


Figure 1: Block diagram of controller

3. Performance test of control system When sensors/actuators are arranged in a collocated way, the so-called positive position feedback (PPF) is extensively used in practice. The modal control structure is described in Figure 2 where the input is η and the output is ξ .

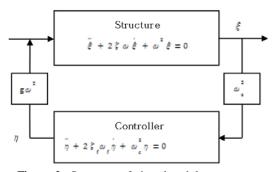


Figure 2: Structure of the closed loop system

In the framework of the PPF design methodology [11,12], the system and control structure for the single-input single-output (SISO) are described modal equations as

$$\ddot{\xi} + 2\xi \omega \dot{\xi} + \omega^2 \xi = g \omega^2 \eta$$
(3)

$$\ddot{\eta} + 2\xi_f \omega_f \eta + \omega_f^2 \eta = \omega_f^2 \xi \tag{4}$$

where

 ξ : modal coordinate of the plate \mathscr{Q} : natural frequency of the plate η : modal coordinate of the controller ζ : damping coeficient of the plate $\omega_{\rm f}$: natural frequency of the controller $\xi_{\rm f}$: damping coeficient of the controller g: feedback control gain

Using the modal dynamic Equations (3) and (4), the transfer functions of the plate in equation (5) and the controller in Equation (6) are expressed as

$$G(s) = \frac{s^2}{s^2 + 2s s^2 + s^2}$$
(5)

$$H(s) = \frac{\nu_f^2}{s^2 + 2t_f \, e^{s} + \nu_f^2} \tag{6}$$

$$s = \frac{2}{T_s} \frac{z - 1}{z + 1}$$
(7)

where, T_s is sampling rate for the compensated circuit (or controller).

The use of a digital control system requires the transformation of the time-continuous equation into its time-discrete form. Substituting (7) into (6) using bilinear transform leads to

$$H(z) = \frac{b_2 z^2 + b_1 z + b_0}{z^2 + a_1 z + a_0}$$
(8)

where,

$$a_{0} = (4/T_{s} - 4\zeta_{f}\omega_{f}/T_{s} + \omega_{f}^{2})/\Omega$$

$$a_{1} = (2\omega_{f}^{2} - 8/T_{s}^{2})/\Omega$$

$$b_{0} = \omega_{f}^{2}/\Omega, \qquad b_{1} = 2\omega_{f}^{2}/\Omega$$

$$b_{2} = \omega_{f}^{2}/\Omega$$

$$\Omega = 4/T_{s}^{2} + 4\zeta_{f}\omega_{f}/T_{s} + \omega_{f}^{2}$$

Then the discrete form for the control algorithm (8) is described in a simple iterative procedure as

$$u_{k} = -a_1 u_{k-1} - a_0 u_{k-2} + b_2 (y_k + 2y_{k-1} + y_{k-2})$$
(9)

where y_k represents the sampled output of the sensor and u_k is the calculated control input.

It is known that feedback control using PPF compensator will accomplish the addition of damping to selective modes [13]. Note that the PPF controller is a highly damped second order low-pass filter that can be tuned to resonate at a flexible structure's natural frequency, much like a tuned mass damper [10]. This compensator could be tuned such that the natural frequency of the compensator (or filter) corresponds the to excitation frequencies, rather than natural frequencies of the structure, and thus provide effective vibration cancellation. At low frequencies this controller adds flexibility and at high frequencies it adds stiffness to the structure. Moreover it also tends to split the vibration modes.

3.1 Sensing test of PZT device

As illustrated in Figure 3, M8557 PZT actuator/sensor [10] is glued to the rectangular aluminum plate supported by the two ends. The plate thickness is 1.5 [*mm*], and its overall width is 210x400 [mm²] in dimensions. The signal of the PZT sensor was measured by the 400 [MHz] digital oscilloscope, and \pm 5 [V] test input signal with 10 [Hz] period was applied using the function generator, and then it was amplified to \pm 100 [V], and the output sensor signal from the PZT was measured as \pm 2 [V] as shown in Figure 4.

The amplified test signal was applied to the PZT actuator, and the voltage signal of the sensor on the vibrated plate was measured. At the rising and falling edge of the step input pulses, the negative and positive sensor signals were measured, respectively. As expected, structural vibration with chattering occurred, and it decreased as time passed. Through the experimental tests, it has been found that the ratio of the actuator input voltage to the sensor output voltage is approximately 1/100.

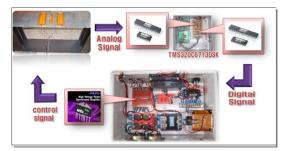


Figure 3: Test bed for vibration control system

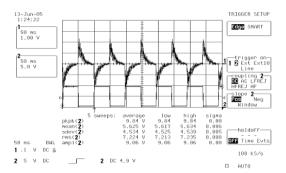
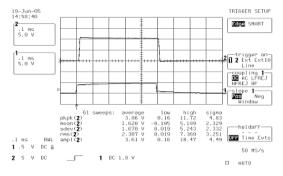


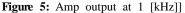
Figure 4: 100 [V] test input with 10Hz (50ms)

3.2 Performance test of drivers

The DSP processor receives the digital signal converted from the analog sensor signal by the A/D converter. According to the control algorithm, the control signal is converted into the analog signal through the D/A converter, which is amplified and put into the PZT actuator. Note that the PZT actuator requires high voltage for actuation. A credible commercial driver PA97 was applied for amplifying the input signal as illustrated in Figures 4 and 5 where the test results of the drivers are presented. For the performance test of the driver, a voltage of 5 [V]

from the function generator was applied to the bridge driver as a step input and amplified output signals were generated in Figure 5 and 6. Also, the processing time from the sensor to the amplified signal of the devices was tested in Figures 5 and 6 where the lower and upper ones represent the sensor signal and the amplified signal, respectively. Figure 6 clearly gives the delay time. 5 [V] clock pulse with 5 [kHz] using the function generator was applied to the A/D converter. Each row line in the upper plot represents 50 [V] and in the lower describes 5 [V]. The processing time between A/D converter and the amplifier took 24.5 [μ s], where A/D conversion, D/A, and amplifier took 10 [μ s], 2 $[\mu s]$, and 12.5 $[\mu s]$, respectively. The amplifier has 8 $[V/\mu s]$ rate such that it took 12.5 $[\mu s]$ in amplifying the voltage up to 100 [V], which is fast enough for active vibration control. Here, the processing time can be reduced by adapting faster peripheral devices.





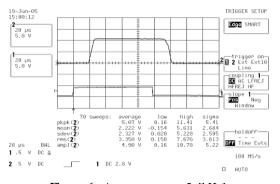


Figure 6: Amp output at 5 [kHz]

3.3 Implementation of the active vibration control

Experiments to test the performance of the developed control system, which is intended to reduce the vibration mode of the aluminium plate have been conducted. Six MFC actuators composed of four M-8528 and two M-8507 were applied on the plate. Through the mode tests of the plate, it is observed that the frequencies of first and second modes are 533 [Hz] and 702 [Hz] using FFT analyzer, respectively. The control system composed of the TMS320c6713 DSP controller and the interrelated Matlab simulink set up as in Figure 7. Using this system, The PPF control algorithm was applied in addition to the FIR filter as in Figure 8. The PPF control algorithm is to reduce the first and second mode of the natural frequency and the FIR filter is to get rid of initial noise signals as shown in Figure 9. By applying the control algorithm using simulink, the vibration modes of the plate are reduced as in Figure 10. In the first mode, about 13.5 dB and in the

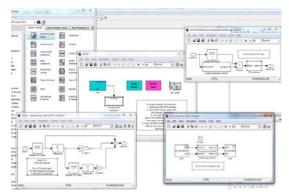


Figure 7: Matlab simulink modeling

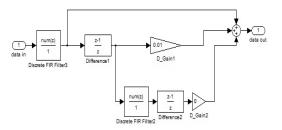


Figure 8: PPF Control & FIR Filter

second mode, 11.8dB were decreased in vibration amplitude.

The TMS320C6713 processor is about 27 times faster than TMS320C30 chip which has been used in many active vibration control tests so far. Experiments were performed by varying the loop time of the controller. Experiments by increasing loop time every 10μ s from 40 to 250 μ s were performed. In Figure 11, results are presented. Consequently, it is identified by the fact that the active control system shows better vibration reduction performance with faster loop times.

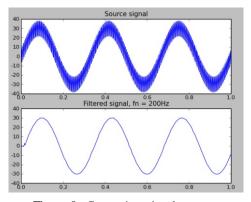


Figure 9: Comparing signal output

In the FIR, linear phase FIR filter algorithm was applied where difference equation of FIR filter [14] is expressed as

$$y(n) = b_0 x(n) + b_1 x(n-1) + \dots + b_{M-1} x(n-(M-1))$$

= $\sum_{k=0}^{M-1} b_k \cdot x(n-k) = \sum_{k=0}^{M-1} h(k) \cdot x(n-k)$ (10)

where $b_k = h(k)$ for $0 \le k \le M - 1$

where b_k is the coefficients and h(k) is impulse response. The FIR filter has linear phase if it satisfies the following condition.

$$h(n) = \pm h(M - 1 - n) \tag{11}$$

for $0 \le n \le M - 1$

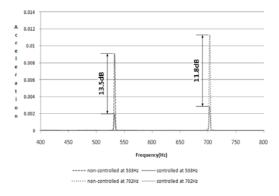


Figure 10: Vibration reduction at the natural frequency

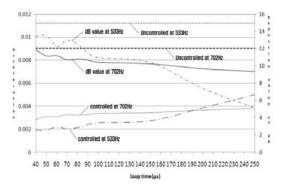


Figure 11: Reduction of vibration by changing loop time

4. Conclusions

In this paper, a new high-speed active control system with the PPF in addition to FIR filter has been developed to suppress the vibration of the rectangular plate. The complete control system using PZT sensor/actuators incorporates the fastest DSP320C6713 microprocessor and peripheral devices such as A/D and D/A converters, and amplifier. Using the developed system, the processing time of the processor and peripheral devices has been tested for the fast speed of the loop times. It is observed that the overall processing time was fast enough to control the vibration mode of the prepared aluminum plate. Additional experiment has been conducted with different loop times. In particular, the PPF control algorithm with the FIR filter has been applied to reduce the specific vibration modes of the plate and noises, and the test results with different loop times have been provided.

As experiment results, firstly, faster loop time shows better results in vibration suppression. Secondly, the proposed PPF control algorithm with FIR filter showed the good vibration suppression performance. Accordingly, it is ensured that the proposed control scheme using DSP320C6713 would be a prospective control system in active vibration control of structures.

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